

### STM32L010xx device errata

#### Applicability

This document applies to the part numbers of STM32L010xx devices and the device variants as stated in this page.

It gives a summary and a description of the device errata, with respect to the device datasheet and reference manual RM0451.

Deviation of the real device behavior from the intended device behavior is considered to be a device limitation. Deviation of the description in the reference manual or the datasheet from the intended device behavior is considered to be a documentation erratum. The term “*errata*” applies both to limitations and documentation errata.

**Table 1. Device summary**

Reference	Part numbers
STM32L010x3	STM32L010D3
STM32L010x4	STM32L010K4, STM32L010F4
STM32L010x6	STM32L010C6
STM32L010x8	STM32L010R8, STM32L010K8
STM32L010xB	STM32L010RB

**Table 2. Device variants**

Reference	Silicon revision codes	
	Device marking <sup>(1)</sup>	REV_ID <sup>(2)</sup>
STM32L010x3/4	Z	0x1008
STM32L010x6	X	0x2018
STM32L010x8	X	0x1008
STM32L010xB	Z	0x2008

1. Refer to the device datasheet for how to identify this code on different types of package.

2. REV\_ID[15:0] bitfield of DBG\_IDCODE register.

## 1 Summary of device errata

The following table gives a quick reference to the STM32L010xx device limitations and their status:

A = limitation present, workaround available

N = limitation present, no workaround available

P = limitation present, partial workaround available

“-” = limitation absent

Applicability of a workaround may depend on specific conditions of target application. Adoption of a workaround may cause restrictions to target application. Workaround for a limitation is deemed partial if it only reduces the rate of occurrence and/or consequences of the limitation, or if it is fully effective for only a subset of instances on the device or in only a subset of operating modes, of the function concerned.

**Table 3. Summary of device limitations**

Function	Section	Limitation	Status			
			Rev. X	Rev. X	Rev. Z	Rev. Z
System	2.1.1	Delay after an RCC peripheral clock enabling	A	A	A	A
	2.1.2	LDM, STM, PUSH and POP not allowed in IOPORT bus	N	N	N	-
	2.1.3	BOOT_MODE bits do not reflect the selected boot mode	-	-	-	N
	2.1.4	Wake-up sequence from Standby mode fails when using more than one wake-up source	A	A	A	A
	2.1.5	LSE crystal oscillator may be disturbed by transitions on PC13	N	N	N	N
	2.1.6	Electrical sensitivity characteristics lower compared to other STM32 devices	-	N	N	-
	2.1.7	Interrupts masked when using dual-bank boot mechanism	-	-	A	-
	2.1.8	NSS pin synchronization required when using bootloader with SPI1 interface on TSSOP14 package	-	-	-	A
GPIO	2.2.1	Writing in byte mode to the GPIOx_OTYPER register does not work	A	-	-	-
	2.2.2	GPIO locking mechanism failed if critical address is read	A	A	A	A
DMA	2.3.1	DMA disable failure and error flag omission upon simultaneous transfer error and global flag clear	A	A	A	A
ADC	2.4.1	Overrun flag is not set if EOC reset coincides with new conversion end	P	P	P	P
	2.4.2	Writing ADC_CFGR1 register while ADEN bit is set resets RES[1:0] bitfield	A	A	A	A
	2.4.3	Out-of-threshold value is not detected in AWD1 Single mode	A	A	A	A
	2.4.4	Writing ADC_CFGR2 register while ADEN bit is set resets CKMODE[1:0] bitfield	A	A	A	A
TIM	2.5.2	Consecutive compare event missed in specific conditions	N	N	N	N
	2.5.3	Output compare clear not working with external counter reset	P	P	P	P
LPTIM	2.6.1	Device may remain stuck in LPTIM interrupt when entering Stop mode	A	A	A	A
	2.6.2	Device may remain stuck in LPTIM interrupt when clearing event flag	P	P	P	P
	2.6.3	LPTIM events and PWM output are delayed by one kernel clock cycle	P	P	P	P
RTC and TAMP	2.7.1	Spurious tamper detection when disabling the tamper channel	P	P	P	-
	2.7.2	RTC calendar registers are not locked properly	A	A	A	A

Function	Section	Limitation	Status			
			Rev. X	Rev. X	Rev. Z	Rev. Z
RTC and TAMP	2.7.3	RTC interrupt can be masked by another RTC interrupt	A	A	A	A
	2.7.4	Calendar initialization may fail in case of consecutive INIT mode entry	A	A	A	A
	2.7.5	Alarm flag may be repeatedly set when the core is stopped in debug	N	N	N	N
	2.7.6	A tamper event fails to trigger timestamp or timestamp overflow events during a few cycles after clearing TSF	N	N	N	N
	2.7.7	Detection of a tamper event occurring before enabling the tamper detection is not supported in edge detection mode	A	A	-	-
I2C	2.8.1	10-bit master mode: new transfer cannot be launched if first part of the address is not acknowledged by the slave	A	A	A	A
	2.8.3	Wrong data sampling when data setup time ( $t_{SU,DAT}$ ) is shorter than one I2C kernel clock period	P	P	P	P
	2.8.4	Spurious bus error detection in master mode	A	A	A	A
	2.8.5	Last-received byte loss in reload mode	P	P	P	P
	2.8.6	Spurious master transfer upon own slave address match	P	P	P	P
	2.8.8	OVR flag not set in underrun condition	N	N	N	N
	2.8.9	Transmission stalled after first byte transfer	A	A	A	A
	2.8.10	SDA held low upon SMBus timeout expiry in slave mode	A	A	A	A
USART	2.9.1	Non-compliant sampling for NACK signal from smartcard	N	N	N	-
	2.9.2	Break request preventing TC flag from being set	A	A	A	-
	2.9.3	RTS is active while RE = 0 or UE = 0	A	A	A	A
	2.9.4	Receiver timeout counter wrong start in two-stop-bit configuration	A	A	A	A
	2.9.5	Data corruption due to noisy receive line	A	A	A	A
	2.9.6	Received data may be corrupted upon clearing the ABREN bit	A	A	A	A
	2.9.7	Noise error flag set while ONEBIT is set	N	N	N	N
	2.9.8	DMA channel 3 (CH3) not functional when USART2_RX used for data reception	-	-	-	A
LPUART	2.10.1	Break request preventing TC flag from being set	A	A	A	-
	2.10.2	RTS is active while RE = 0 or UE = 0	A	A	A	-
	2.10.3	Receiver timeout counter wrong start in two-stop-bit configuration	A	A	A	-
	2.10.4	Data corruption due to noisy receive line	A	A	A	-
	2.10.5	Possible LPUART transmitter issue when using low BRR[15:0] value	P	P	P	-
	2.10.6	DMA channel 5 (CH5) not functional when LPUART1_RX used for data reception	-	-	-	A
SPI/I2S	2.11.1	BSY bit may stay high when SPI is disabled	A	A	A	A
	2.11.2	BSY bit may stay high at the end of data transfer in slave mode	A	A	A	A
	2.11.3	Corrupted last bit of data and/or CRC, received in master mode with delayed SCK feedback	A	A	A	-
	2.11.4	Corrupted last bit of data and/or CRC, received in master mode with delayed SCK feedback	-	-	-	A
	2.11.5	SPI CRC corruption upon DMA transaction completion by another peripheral	P	-	P	-

Function	Section	Limitation	Status			
			Rev. X	Rev. X	Rev. Z	Rev. Z
SPI/I2S	2.11.6	In I <sup>2</sup> S slave mode, enabling I2S while WS is active causes desynchronization	A	-	A	-
	2.11.7	Wrong CRC in full-duplex mode handled by DMA with imbalanced setting of data counters	A	A	A	A
	2.11.9	Anticipated communication upon SPI transit from slave receiver to master	A	A	A	A
USB	2.12.2	ESOF interrupt timing desynchronized after resume signaling	A	-	A	-
	2.12.3	Incorrect CRC16 in the memory buffer	N	-	N	-
	2.12.4	Buffer description table update completes after CTR interrupt triggers	A	-	A	-
	2.12.5	USB BCD functionality limited below -20 °C	N	-	N	-

The following table gives a quick reference to the documentation errata.

**Table 4. Summary of device documentation errata**

Function	Section	Documentation erratum
DMA	2.3.2	Byte and half-word accesses not supported
TIM	2.5.1	TRGO and TRGO2 trigger output failure
I2C	2.8.2	Wrong behavior in Stop mode
	2.8.7	START bit is cleared upon setting ADDRCONF, not upon address match
SPI/I2S	2.11.8	CRC error in SPI slave mode if internal NSS changes before CRC transfer
USB	2.12.1	Possible packet memory overrun/underrun at low APB frequency

## 2 Description of device errata

The following sections describe the errata of the applicable devices with Arm® core and provide workarounds if available. They are grouped by device functions.

*Note: Arm is a registered trademark of Arm Limited (or its subsidiaries) in the US and/or elsewhere.*

**arm**

### 2.1 System

#### 2.1.1 Delay after an RCC peripheral clock enabling

##### Description

A delay between an RCC peripheral clock enable and the effective peripheral enabling should be taken into account in order to manage the peripheral read/write from/to registers.

This delay depends on the peripheral mapping:

- If the peripheral is mapped on AHB: the delay should be equal to 1 AHB clock cycle after the clock enable bit is set in the hardware register.  
For I/O peripheral, the delay should be equal to 1 AHB clock cycle after the clock enable bit is set in the hardware register (only applicable to write accesses).
- If the peripheral is mapped on APB: No delay is necessary (no limitation).

##### Workaround

- Enable the peripheral clock some time before the peripheral read/write register is required.
- For AHB peripheral (including I/O), insert a dummy read operation to the corresponding register.

#### 2.1.2 LDM, STM, PUSH and POP not allowed in IOPORT bus

##### Description

The instructions Load Multiple (LM), Store Multiple (STM), PUSH and POP fail when the address points to the IOPORT bus memory area (address range = 0x5XXX XXXX).

##### Workaround

None.

#### 2.1.3 BOOT\_MODE bits do not reflect the selected boot mode

##### Description

The BOOT\_MODE[1:0] bits of the SYSCFG\_CFGR1 register remain set to '0' while they should reflect the boot mode selected by the boot pins.

##### Workaround

None.

#### 2.1.4 Wake-up sequence from Standby mode fails when using more than one wake-up source

##### Description

The various wake-up sources are logically OR-ed in front of the rising-edge detector which generates the wake-up flag (WUF). The WUF flag needs to be cleared prior to the Standby mode entry, otherwise the microcontroller wakes up immediately.

If one of the configured wake-up sources is kept high during the WUF flag clearing (by setting the CWUF bit), it may mask further wake-up events on the input of the edge detector. As a consequence, the microcontroller cannot wake up from Standby mode.

#### Workaround

Apply the following sequence before entering Standby mode:

1. Disable all used wake-up sources.
2. Clear all related wake-up flags.
3. Re-enable all used wake-up sources.
4. Enter Standby mode.

**Note:** *When applying this measure, if one of the wake-up sources is still kept high, be aware that the microcontroller enters Standby mode but then wakes up immediately, generating the power reset.*

### 2.1.5 LSE crystal oscillator may be disturbed by transitions on PC13

#### Description

On LQFP and UFQFPN packages, the LSE crystal oscillator clock frequency can be incorrect when PC13 is toggling in input or output (for example when used for RTC\_OUT1). The external clock input (LSE bypass) is not impacted by this limitation. The WLCSP and UFBGA packages are not impacted by this limitation.

Avoid toggling PC13 when LSE is used on LQFP and UFQFPN packages.

#### Workaround

None.

### 2.1.6 Electrical sensitivity characteristics lower compared to other STM32 devices

#### Description

The ESD absolute maximum ratings are lower compared to some other STM32 devices:

- Electrostatic discharge voltage human body model ( $V_{ESD(HBM)}$ ) class is 1C instead of 2 and the maximum value is 1000 V instead of 2000 V.
- Electrostatic discharge voltage charge device model ( $V_{ESD(CDM)}$ ) class is C3 and the maximum value is 250 V instead of 500 V.

#### Workaround

None.

### 2.1.7 Interrupts masked when using dual-bank boot mechanism

#### Description

When the dual-bank boot mechanism is enabled, either by setting BFB2 bit in FLASH\_OTPR register, the microcontroller starts executing the user code with bit 0 of the core priority mask register (PRIMASK) set. As a result, the interrupts are not serviced.

#### Workaround

Set bit 0 of PRIMASK register before activating any interrupt. See code below for an example:

```
__ASM ("CPSIE i");
```

### 2.1.8 NSS pin synchronization required when using bootloader with SPI1 interface on TSSOP14 package

#### Description

When using the embedded bootloader with SPI1 interface on devices in TSSOP14 package, if the NSS pin is grounded (default status after device reset), SPI communications are not synchronized and the bootloader does not work.

To properly synchronize the SPI interface, an NSS pin falling edge is required before initiating communications.

#### Workaround

On devices in TSSOP14, toggle with NSS pin (PA14) after device reset.

## 2.2 GPIO

### 2.2.1 Writing in byte mode to the GPIOx\_OTYPER register does not work

#### Description

The OTYPER[15:8] bits in GPIOx\_OTYPER register cannot be written in byte mode. This is valid for A, B, C, D and H ports.

However, the following operations are possible:

- OTYPER[15:8] bits can be written in half-word and word mode
- OTYPER[7:0] bits can be written in byte, half-word or word mode

#### Workaround

Program GPIOx\_OTYPER bits in half-word or word mode.

### 2.2.2 GPIO locking mechanism failed if critical address is read

#### Description

The GPIO locking sequence requires predefined steps of writing and reading operations over the GPIOx\_LCKR register. Reading from an address (SRAM/flash/peripherals) where the 10 least significant bits corresponds to the GPIOx\_LCKR address (0x5000 XX1C) is interpreted by GPIO locking state machine as a dummy read from the GPIOx\_LCKR register. The GPIO locking sequence fails when an additional dummy read from the GPIOx\_LCKR register is inserted, resetting the locking sequence. The GPIO locking state machine is only sensitive to reading done by CPU and not by DMA.

#### Workaround

The GPIO locking sequence code must not use variables stored at SRAM addresses which 10 least significant bits correspond to 0x01C (each 1kB is a critical address). Avoid using variables on the stack because critical addresses can be present on the stack area in the locking sequence code. Disable interrupts during the GPIO locking sequence because interrupt service routine may perform reading from critical address. The locking sequence routine must be placed in a memory section outside of critical addresses.

Find recommended code for GPIO locking sequence below. It only uses register access with disabled interrupts in critical code part. The correct code placement (outside critical addresses) is solved here by repeating the locking sequence in case of error, so that at least one locking code is at correct placement (linker independent solution).

```
/* (1) Save interrupts enable state */
/* (2) Disable interrupts */
/* (3) Write LCKK bit to 1 and set the pin bits to lock */
/* (4) Write LCKK bit to 0 and set the pin bits to lock */
/* (5) Write LCKK bit to 1 and set the pin bits to lock */
/* (6) Restore interrupts enable state */
/* (7) Read the Lock register */
/* (8) Check the Lock register for lock (in case of error perform lock second time) */

register uint32_t tmp1; /* place critical variable into register to not use SRAM */
```

```

register uint32_t tmp2; /* place critical variable into register to not use SRAM */
uint32_t irgenabled;

tmp1 = GPIO_Pin;
tmp2 = GPIO_LCKR_LCKK | GPIO_Pin;

...
irgenabled = __get_PRIMASK(); /* (1) */ /* remember interrupts enable state */
__disable_irq(); /* (2) */
GPIOA->LCKR = tmp2; /* (3) */
GPIOA->LCKR = tmp1; /* (4) */
GPIOA->LCKR = tmp2; /* (5) */
if (!irgenabled) __enable_irq(); /* (6) */
GPIOA->LCKR; /* (7) */
if ((GPIOA->LCKR & GPIO_LCKR_LCKK) == 0) /* (8) */ /* check if locking failed */
{
  __disable_irq(); /* (2) */
  GPIOA->LCKR = tmp2; /* (3) */
  GPIOA->LCKR = tmp1; /* (4) */
  GPIOA->LCKR = tmp2; /* (5) */
  if (!irgenabled) __enable_irq(); /* (6) */
  GPIOA->LCKR; /* (7) */
  if ((GPIOA->LCKR & GPIO_LCKR_LCKK) == 0) /* (8) */
  {
    /* Manage error - optional */
  }
}

```

## 2.3 DMA

### 2.3.1 DMA disable failure and error flag omission upon simultaneous transfer error and global flag clear

#### Description

Upon a data transfer error in a DMA channel x, both the specific TEIFx and the global GIFx flags are raised and the channel x is normally automatically disabled. However, if in the same clock cycle the software clears the GIFx flag (by setting the CGIFx bit of the DMA\_IFCR register), the automatic channel disable fails and the TEIFx flag is not raised.

This issue does not occur with ST's HAL software that does not use and clear the GIFx flag when the channel is active.

#### Workaround

Do not clear GIFx flags when the channel is active. Instead, use HTIFx, TCIFx, and TEIFx specific event flags and their corresponding clear bits.

### 2.3.2 Byte and half-word accesses not supported

#### Description

Some reference manual revisions may wrongly state that the DMA registers are byte- and half-word-accessible. Instead, the DMA registers must always be accessed through aligned 32-bit words. Byte or half-word write accesses cause an erroneous behavior.

ST's low-level driver and HAL software only use aligned 32-bit accesses to the DMA registers.

This is a description inaccuracy issue rather than a product limitation.

#### Workaround

No application workaround is required.



## 2.4 ADC

### 2.4.1 Overrun flag is not set if EOC reset coincides with new conversion end

#### Description

If the EOC flag is cleared by an ADC\_DR register read operation or by software during the same APB cycle in which the data from a new conversion are written in the ADC\_DR register, the overrun event duly occurs (which results in the loss of either current or new data) but the overrun flag (OVR) may stay low.

#### Workaround

Clear the EOC flag, by performing an ADC\_DR read operation or by software within less than one ADC conversion cycle period from the last conversion cycle end, in order to avoid the coincidence with the end of the new conversion cycle.

### 2.4.2 Writing ADC\_CFGR1 register while ADEN bit is set resets RES[1:0] bitfield

#### Description

Modifying the ADC\_CFGR1 register while ADC is enabled (ADEN set in ADC\_CR) resets RES[1:0] to 00 whatever the bitfield previous value.

#### Workaround

Apply the following sequence:

1. Set ADDIS to disable the ADC, and wait until ADEN is cleared.
2. Program the ADC\_CFGR1 register according to the application requirements.
3. Set ADEN bit.

### 2.4.3 Out-of-threshold value is not detected in AWD1 Single mode

#### Description

AWD1 analog watchdog does not detect that the result of a converted channel has reached the programmed threshold when the ADC operates in Single mode, performs a sequence of conversions, and one of the converted channels other than the first one is monitored by the AWD1 analog watchdog.

#### Workaround

Apply one of the following measures:

- Use a conversion sequence of one single channel.
- Configure the monitored channel as the first one of the sequence.

### 2.4.4 Writing ADC\_CFGR2 register while ADEN bit is set resets CKMODE[1:0] bitfield

#### Description

Modifying the ADC\_CFGR2 register while ADC is enabled (ADEN set in ADC\_CR) resets CKMODE[1:0] to 00 whatever the bitfield previous value.

#### Workaround

Apply the following sequence:

1. Set ADDIS to disable the ADC, and wait until ADEN is cleared.
2. Program the ADC\_CFGR2 register according to the application requirements.
3. Set ADEN bit.

## 2.5 TIM

### 2.5.1 TRGO and TRGO2 trigger output failure

#### Description

Some reference manual revisions may omit the following information.

The timers can be linked using ITRx inputs and TRGOx outputs. Additionally, the TRGOx outputs can be used as triggers for other peripherals (for example ADC). Since this circuitry is based on pulse generation, care must be taken when initializing master and slave peripherals or when using different master/slave clock frequencies:

- If the master timer generates a trigger output pulse on TRGOx prior to have the destination peripheral clock enabled, the triggering system may fail.
- If the frequency of the destination peripheral is modified on-the-fly (clock prescaler modification), the triggering system may fail.

As a conclusion, the clock of the slave timer or slave peripheral must be enabled prior to receiving events from the master timer, and must not be changed on-the-fly while triggers are being received from the master timer.

This is a documentation issue rather than a product limitation.

#### Workaround

No application workaround is required or applicable as long as the application handles the clock as indicated.

### 2.5.2 Consecutive compare event missed in specific conditions

#### Description

Every match of the counter (CNT) value with the compare register (CCR) value is expected to trigger a compare event. However, if such matches occur in two consecutive counter clock cycles (as consequence of the CCR value change between the two cycles), the second compare event is missed for the following CCR value changes:

- in edge-aligned mode, from ARR to 0:
  - first compare event:  $CNT = CCR = ARR$
  - second (missed) compare event:  $CNT = CCR = 0$
- in center-aligned mode while up-counting, from ARR-1 to ARR (possibly a new ARR value if the period is also changed) at the crest (that is, when  $TIMx\_RCR = 0$ ):
  - first compare event:  $CNT = CCR = (ARR-1)$
  - second (missed) compare event:  $CNT = CCR = ARR$
- in center-aligned mode while down-counting, from 1 to 0 at the valley (that is, when  $TIMx\_RCR = 0$ ):
  - first compare event:  $CNT = CCR = 1$
  - second (missed) compare event:  $CNT = CCR = 0$

This typically corresponds to an abrupt change of compare value aiming at creating a timer clock single-cycle-wide pulse in toggle mode.

As a consequence:

- In toggle mode, the output only toggles once per counter period (squared waveform), whereas it is expected to toggle twice within two consecutive counter cycles (and so exhibit a short pulse per counter period).
- In center mode, the compare interrupt flag does not rise and the interrupt is not generated.

*Note:* The timer output operates as expected in modes other than the toggle mode.

#### Workaround

None.

### 2.5.3 Output compare clear not working with external counter reset

#### Description

The output compare clear event (ocref\_clr) is not correctly generated when the timer is configured in the following slave modes: Reset mode, Combined reset + trigger mode, and Combined gated + reset mode.

The PWM output remains inactive during one extra PWM cycle if the following sequence occurs:

1. The output is cleared by the ocref\_clr event.
2. The timer reset occurs before the programmed compare event.

#### Workaround

Apply one of the following measures:

- Use BKIN (or BKIN2 if available) input for clearing the output, selecting the Automatic output enable mode (AOE = 1).
- Mask the timer reset during the PWM ON time to prevent it from occurring before the compare event (for example with a spare timer compare channel open-drain output connected with the reset signal, pulling the timer reset line down).

## 2.6 LPTIM

### 2.6.1 Device may remain stuck in LPTIM interrupt when entering Stop mode

#### Description

This limitation occurs when disabling the low-power timer (LPTIM).

When the user application clears the ENABLE bit in the LPTIM\_CR register within a small time window around one LPTIM interrupt occurrence, then the LPTIM interrupt signal used to wake up the device from Stop mode may be frozen in active state. Consequently, when trying to enter Stop mode, this limitation prevents the device from entering low-power mode and the firmware remains stuck in the LPTIM interrupt routine.

This limitation applies to all Stop modes and to all instances of the LPTIM. Note that the occurrence of this issue is very low.

#### Workaround

In order to disable a low power timer (LPTIMx) peripheral, do not clear its ENABLE bit in its respective LPTIM\_CR register. Instead, reset the whole LPTIMx peripheral via the RCC controller by setting and resetting its respective LPTIMxRST bit in the relevant RCC register.

### 2.6.2 Device may remain stuck in LPTIM interrupt when clearing event flag

#### Description

This limitation occurs when the LPTIM is configured in interrupt mode (at least one interrupt is enabled) and the software clears any flag in LPTIM\_ISR register by writing its corresponding bit in LPTIM\_ICR register. If the interrupt status flag corresponding to a disabled interrupt is cleared simultaneously with a new event detection, the set and clear commands might reach the APB domain at the same time, leading to an asynchronous interrupt signal permanently stuck high.

This issue can occur either during an interrupt subroutine execution (where the flag clearing is usually done), or outside an interrupt subroutine.

Consequently, the firmware remains stuck in the LPTIM interrupt routine, and the device cannot enter Stop mode.

### Workaround

To avoid this issue, it is strongly advised to follow the recommendations listed below:

- Clear the flag only when its corresponding interrupt is enabled in the interrupt enable register.
- If for specific reasons, it is required to clear some flags that have corresponding interrupt lines disabled in the interrupt enable register, it is recommended to clear them during the current subroutine prior to those which have corresponding interrupt line enabled in the interrupt enable register.
- Flags must not be cleared outside the interrupt subroutine.

**Note:** *The standard clear sequence implemented in the HAL\_LPTIM\_IRQHandler in the STM32Cube is considered as the proper clear sequence.*

## 2.6.3 LPTIM events and PWM output are delayed by one kernel clock cycle

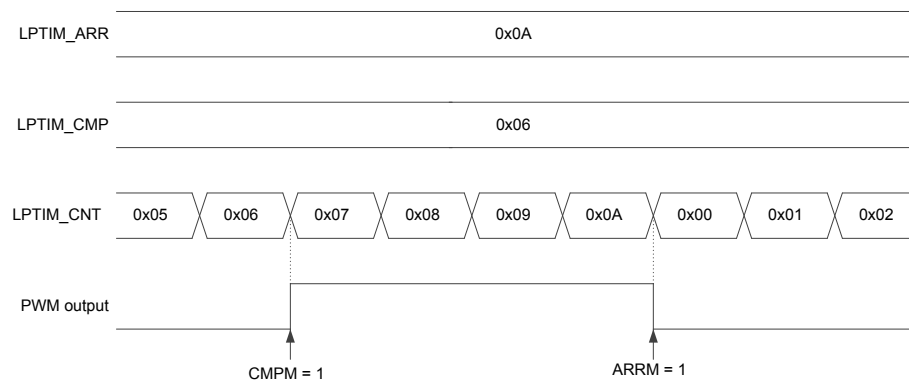
### Description

The compare match event (CMPM), auto reload match event (ARRM), PWM output level and interrupts are updated with a delay of one kernel clock cycle.

Consequently, it is not possible to generate PWM with a duty cycle of 0% or 100%.

The following waveform gives the example of PWM output mode and the effect of the delay:

**Figure 1. Example of PWM output mode**



### Workaround

Set the compare value to the desired value minus 1. For instance in order to generate a compare match when LPTM\_CNT = 0x08, set the compare value to 0x07.

## 2.7 RTC and TAMP

### 2.7.1 Spurious tamper detection when disabling the tamper channel

#### Description

If the tamper detection is configured for detecting on the falling edge event (TAMPFLT = 00 and TAMPxTRG = 1) and if the tamper event detection is disabled when the tamper pin is at high level, a false tamper event is detected, which may result in the erasure of backup registers.

#### Workaround

None for the false detection of tamper event. The erasure of the backup registers can be avoided by setting the TAMPxNOERASE bit before clearing the TAMPxE bit, in two separate RTC\_TAMPCR write accesses.

## 2.7.2 RTC calendar registers are not locked properly

### Description

When reading the calendar registers with BYPSHAD = 0, the RTC\_TR and RTC\_DR registers may not be locked after reading the RTC\_SSR register. This happens if the read operation is initiated one APB clock period before the shadow registers are updated. This can result in a non-consistency of the three registers. Similarly, the RTC\_DR register can be updated after reading the RTC\_TR register instead of being locked.

### Workaround

Apply one of the following measures:

- Use BYPSHAD = 1 mode (bypass shadow registers), or
- If BYPSHAD = 0, read SSR again after reading SSR/TR/DR to confirm that SSR is still the same, otherwise read the values again.

## 2.7.3 RTC interrupt can be masked by another RTC interrupt

### Description

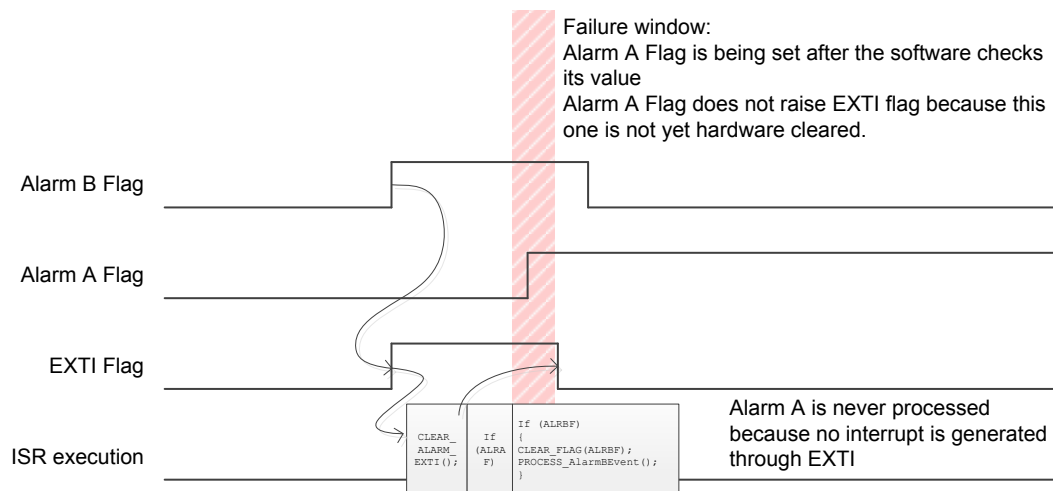
One RTC interrupt request can mask another RTC interrupt request if they share the same EXTI configurable line. For example, interrupt requests from Alarm A and Alarm B or those from tamper and timestamp events are OR-ed to the same EXTI line (refer to the *EXTI line connections* table in the *Extended interrupt and event controller (EXTI)* section of the reference manual).

The following code example and figure illustrate the failure mechanism: The Alarm A event is lost (fails to generate interrupt) as it occurs in the failure window, that is, after checking the Alarm A event flag but before the effective clear of the EXTI interrupt flag by hardware. The effective clear of the EXTI interrupt flag is delayed with respect to the software instruction to clear it.

Alarm interrupt service routine:

```
void RTC_Alarm_IRQHandler(void)
{
    CLEAR_ALARM_EXTI(); /* Clear the EXTI line flag for RTC alarms*/
    If(ALRAF) /* Check if Alarm A triggered ISR */
    {
        CLEAR_FLAG(ALRAF); /* Clear the Alarm A interrupt pending bit */
        PROCESS_AlarmAEvent(); /* Process Alarm A event */
    }
    If(ALRBF) /* Check if Alarm B triggered ISR */
    {
        CLEAR_FLAG(ALRBF); /* Clear the Alarm B interrupt pending bit */
        PROCESS_AlarmBEvent(); /* Process Alarm B event */
    }
}
```

Figure 2. Masked RTC interrupt



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### Workaround

In the interrupt service routine, apply three consecutive event flag checks - source one, source two, and source one again, as in the following code example:

```

void RTC_Alarm_IRQHandler(void)
{
    CLEAR_ALARM_EXTI(); /* Clear the EXTI's line Flag for RTC Alarm */
    If(ALRAF) /* Check if AlarmA triggered ISR */
    {
        CLEAR_FLAG(ALRAF); /* Clear the AlarmA interrupt pending bit */
        PROCESS_AlarmAEvent(); /* Process AlarmA Event */
    }
    If(ALRBF) /* Check if AlarmB triggered ISR */
    {
        CLEAR_FLAG(ALRBF); /* Clear the AlarmB interrupt pending bit */
        PROCESS_AlarmBEvent(); /* Process AlarmB Event */
    }
    If(ALRAF) /* Check if AlarmA triggered ISR */
    {
        CLEAR_FLAG(ALRAF); /* Clear the AlarmA interrupt pending bit */
        PROCESS_AlarmAEvent(); /* Process AlarmA Event */
    }
}
    
```

## 2.7.4 Calendar initialization may fail in case of consecutive INIT mode entry

### Description

If the INIT bit of the RTC\_ISR register is set between one and two RTCCLK cycles after being cleared, the INITF flag is set immediately instead of waiting for synchronization delay (which should be between one and two RTCCLK cycles), and the initialization of registers may fail.

Depending on the INIT bit clearing and setting instants versus the RTCCLK edges, it can happen that, after being immediately set, the INITF flag is cleared during one RTCCLK period then set again. As writes to calendar registers are ignored when INITF is low, a write during this critical period might result in the corruption of one or more calendar registers.

### Workaround

After exiting the initialization mode, clear the BYPSHAD bit (if set) then wait for RSF to rise, before entering the initialization mode again.

**Note:** *It is recommended to write all registers in a single initialization session to avoid accumulating synchronization delays.*

### 2.7.5 Alarm flag may be repeatedly set when the core is stopped in debug

#### Description

When the core is stopped in debug mode, the clock is supplied to subsecond RTC alarm downcounter even when the device is configured to stop the RTC in debug.

As a consequence, when the subsecond counter is used for alarm condition (the MASKSS[3:0] bitfield of the RTC\_ALRMASRR and/or RTC\_ALRMBSSR register set to a non-zero value) and the alarm condition is met just before entering a breakpoint or printf, the ALRAF and/or ALRBF flag of the RTC\_SR register is repeatedly set by hardware during the breakpoint or printf, which makes any attempt to clear the flag(s) ineffective.

#### Workaround

None.

### 2.7.6 A tamper event fails to trigger timestamp or timestamp overflow events during a few cycles after clearing TSF

#### Description

With the timestamp on tamper event enabled (TAMPTS bit of the RTC\_CR register set), a tamper event is ignored if it occurs:

- within four APB clock cycles after setting the CTSF bit of the RTC\_SCR register to clear the TSF flag, while the TSF flag is not yet effectively cleared (it fails to set the TSOVF flag)
- within two ck\_apre cycles after setting the CTSF bit of the RTC\_SCR register to clear the TSF flag, when the TSF flag is effectively cleared (it fails to set the TSF flag and timestamp the calendar registers)

#### Workaround

None.

### 2.7.7 Detection of a tamper event occurring before enabling the tamper detection is not supported in edge detection mode

#### Description

When the tamper detection is enabled in edge detection mode (TAMPFLT=00):

- When TAMPxTRG=0 (rising edge detection): if the tamper input is already high before enabling the tamper detection, the tamper event may or may not be detected when enabling the tamper detection. The probability to detect it increases with the APB frequency.
- When TAMPxTRG=1 (falling edge detection): if the tamper input is already low before enabling the tamper detection, the tamper event is not detected when enabling the tamper detection.

#### Workaround

The I/O state should be checked by software in the GPIO registers, just after enabling the tamper detection and before writing sensitive values in the backup registers, in order to ensure that no active edge occurred before enabling the tamper event detection.

## 2.8 I2C

### 2.8.1 10-bit master mode: new transfer cannot be launched if first part of the address is not acknowledged by the slave

#### Description

An I<sup>2</sup>C-bus master generates STOP condition upon non-acknowledge of I<sup>2</sup>C address that it sends. This applies to 7-bit address as well as to each byte of 10-bit address.

When the MCU set as I<sup>2</sup>C-bus master transmits a 10-bit address of which the first byte (5-bit header + 2 MSBs of the address + direction bit) is not acknowledged, the MCU duly generates STOP condition but it then cannot start any new I<sup>2</sup>C-bus transfer. In this spurious state, the NACKF flag of the I2C\_ISR register and the START bit of the I2C\_CR2 register are both set, while the START bit should normally be cleared.

#### Workaround

In 10-bit-address master mode, if both NACKF flag and START bit get simultaneously set, proceed as follows:

1. Wait for the STOP condition detection (STOPF = 1 in I2C\_ISR register).
2. Disable the I2C peripheral.
3. Wait for a minimum of three APB cycles.
4. Enable the I2C peripheral again.

### 2.8.2 Wrong behavior in Stop mode

#### Description

The correct use of the I2C peripheral is to disable it (PE = 0) before entering Stop mode, and re-enable it when back in Run mode.

Some reference manual revisions may omit this information.

Failure to respect the above while the MCU operating as slave or as master in multi-master topology enters Stop mode during a transfer ongoing on the I<sup>2</sup>C-bus may lead to the following:

1. BUSY flag is wrongly set when the MCU exits Stop mode. This prevents from initiating a transfer in master mode, as the START condition cannot be sent when BUSY is set.
2. If clock stretching is enabled (NOSTRETCH = 0), the SCL line is pulled low by I2C and the transfer stalled as long as the MCU remains in Stop mode.

The occurrence of such condition depends on the timing configuration, peripheral clock frequency, and I<sup>2</sup>C-bus frequency.

This is a description inaccuracy issue rather than a product limitation.

#### Workaround

No application workaround is required.

### 2.8.3 Wrong data sampling when data setup time ( $t_{\text{SU;DAT}}$ ) is shorter than one I2C kernel clock period

#### Description

The I<sup>2</sup>C-bus specification and user manual specify a minimum data setup time ( $t_{\text{SU;DAT}}$ ) as:

- 250 ns in Standard mode
- 100 ns in Fast mode
- 50 ns in Fast mode Plus

The device does not correctly sample the I<sup>2</sup>C-bus SDA line when  $t_{\text{SU;DAT}}$  is smaller than one I2C kernel clock (I<sup>2</sup>C-bus peripheral clock) period: the previous SDA value is sampled instead of the current one. This can result in a wrong receipt of slave address, data byte, or acknowledge bit.



### Workaround

Increase the I2C kernel clock frequency to get I2C kernel clock period within the transmitter minimum data setup time. Alternatively, increase transmitter's minimum data setup time. If the transmitter setup time minimum value corresponds to the minimum value provided in the I<sup>2</sup>C-bus standard, the minimum I2CCLK frequencies are as follows:

- In Standard mode, if the transmitter minimum setup time is 250 ns, the I2CCLK frequency must be at least 4 MHz.
- In Fast mode, if the transmitter minimum setup time is 100 ns, the I2CCLK frequency must be at least 10 MHz.
- In Fast-mode Plus, if the transmitter minimum setup time is 50 ns, the I2CCLK frequency must be at least 20 MHz.

## 2.8.4 Spurious bus error detection in master mode

### Description

In master mode, a bus error can be detected spuriously, with the consequence of setting the BERR flag of the I2C\_SR register and generating bus error interrupt if such interrupt is enabled. Detection of bus error has no effect on the I<sup>2</sup>C-bus transfer in master mode and any such transfer continues normally.

### Workaround

If a bus error interrupt is generated in master mode, the BERR flag must be cleared by software. No other action is required and the ongoing transfer can be handled normally.

## 2.8.5 Last-received byte loss in reload mode

### Description

If in master receiver mode or slave receive mode with SBC = 1 the following conditions are all met:

- I<sup>2</sup>C-bus stretching is enabled (NOSTRETCH = 0)
- RELOAD bit of the I2C\_CR2 register is set
- NBYTES bitfield of the I2C\_CR2 register is set to N greater than 1
- byte N is received on the I<sup>2</sup>C-bus, raising the TCR flag
- N - 1 byte is not yet read out from the data register at the instant TCR is raised,

then the SCL line is pulled low (I<sup>2</sup>C-bus clock stretching) and the transfer of the byte N from the shift register to the data register inhibited until the byte N-1 is read and NBYTES bitfield reloaded with a new value, the latter of which also clears the TCR flag. As a consequence, the software cannot get the byte N and use its content before setting the new value into the NBYTES field.

### Workaround

- In master mode or in slave mode with SBC = 1, use the reload mode with NBYTES = 1.
- In master receiver mode, if the number of bytes to transfer is greater than 255, do not use the reload mode. Instead, split the transfer into sections not exceeding 255 bytes and separate them with repeated START conditions.
- Make sure, for example through the use of DMA, that the byte N - 1 is always read before the TCR flag is raised.

The last workaround in the list must be evaluated carefully for each application as the timing depends on factors such as the bus speed, interrupt management, software processing latencies, and DMA channel priority.

## 2.8.6 Spurious master transfer upon own slave address match

### Description

When the device is configured to operate at the same time as master and slave (in a multi-master I<sup>2</sup>C-bus application), a spurious master transfer may occur under the following condition:

- Another master on the bus is in process of sending the slave address of the device (the bus is busy).
- The device initiates a master transfer by bit set before the slave address match event (the ADDR flag set in the I2C\_ISR register) occurs.
- After the ADDR flag is set:
  - the device does not write I2C\_CR2 before clearing the ADDR flag, or
  - the device writes I2C\_CR2 earlier than three I2C kernel clock cycles before clearing the ADDR flag

In these circumstances, even though the START bit is automatically cleared by the circuitry handling the ADDR flag, the device spuriously proceeds to the master transfer as soon as the bus becomes free. The transfer configuration depends on the content of the I2C\_CR2 register when the master transfer starts. Moreover, if the I2C\_CR2 is written less than three kernel clocks before the ADDR flag is cleared, the I2C peripheral may fall into an unpredictable state.

### Workaround

Upon the address match event (ADDR flag set), apply the following sequence.

Normal mode (SBC = 0):

1. Set the ADDRCONF bit.
2. Before Stop condition occurs on the bus, write I2C\_CR2 with the START bit low.

Slave byte control mode (SBC = 1):

1. Write I2C\_CR2 with the slave transfer configuration and the START bit low.
2. Wait for longer than three I2C kernel clock cycles.
3. Set the ADDRCONF bit.
4. Before Stop condition occurs on the bus, write I2C\_CR2 again with its current value.

The time for the software application to write the I2C\_CR2 register before the Stop condition is limited, as the clock stretching (if enabled), is aborted when clearing the ADDR flag.

Polling the BUSY flag before requesting the master transfer is not a reliable workaround as the bus may become busy between the BUSY flag check and the write into the I2C\_CR2 register with the START bit set.

## 2.8.7 START bit is cleared upon setting ADDRCONF, not upon address match

### Description

Some reference manual revisions may state that the START bit of the I2C\_CR2 register is cleared upon slave address match event.

Instead, the START bit is cleared upon setting, by software, the ADDRCONF bit of the I2C\_ICR register, which does not guarantee the abort of master transfer request when the device is being addressed as slave. This product limitation and its workaround are the subject of a separate erratum.

### Workaround

No application workaround is required for this description inaccuracy issue.

## 2.8.8 OVR flag not set in underrun condition

### Description

In slave transmission with clock stretching disabled (NOSTRETCH = 1 in the I2C\_CR1 register), an underrun condition occurs if the current byte transmission is completed on the I<sup>2</sup>C bus, and the next data is not yet written in the TXDATA[7:0] bitfield. In this condition, the device is expected to set the OVR flag of the I2C\_ISR register and send 0xFF on the bus.

However, if the I2C\_TXDR is written within the interval between two I2C kernel clock cycles before and three APB clock cycles after the start of the next data transmission, the OVR flag is not set, although the transmitted value is 0xFF.

#### Workaround

None.

### 2.8.9 Transmission stalled after first byte transfer

#### Description

When the first byte to transmit is not prepared in the TXDATA register, two bytes are required successively, through TXIS status flag setting or through a DMA request. If the first of the two bytes is written in the I2C\_TXDR register in less than two I2C kernel clock cycles after the TXIS/DMA request, and the ratio between APB clock and I2C kernel clock frequencies is between 1.5 and 3, the second byte written in the I2C\_TXDR is not internally detected. This causes a state in which the I2C peripheral is stalled in master mode or in slave mode, with clock stretching enabled (NOSTRETCH = 0). This state can only be released by disabling the peripheral (PE = 0) or by resetting it.

#### Workaround

Apply one of the following measures:

- Write the first data in I2C\_TXDR before the transmission starts.
- Set the APB clock frequency so that its ratio with respect to the I2C kernel clock frequency is lower than 1.5 or higher than 3.

### 2.8.10 SDA held low upon SMBus timeout expiry in slave mode

#### Description

For the slave mode, the SMBus specification defines  $t_{\text{TIMEOUT}}$  (detect clock low timeout) and  $t_{\text{LOW:SEXT}}$  (cumulative clock low extend time) timeouts. When one of them expires while the I2C peripheral in slave mode drives SDA low to acknowledge either its address or a data transmitted by the master, the device is expected to report such an expiry and release the SDA line.

However, although the device duly reports the timeout expiry, it fails to release SDA. This stalls the I<sup>2</sup>C bus and prevents the master from generating RESTART or STOP condition.

#### Workaround

When a timeout is reported in slave mode (TIMEOUT bit of the I2C\_ISR register is set), apply this sequence:

1. Wait until the frame is expected to end.
2. Read the STOPF bit of the I2C\_ISR register. If it is low, reset the I2C kernel by clearing the PE bit of the I2C\_CR1 register.
3. Wait for at least three APB clock cycles before enabling again the I2C peripheral.

## 2.9 USART

### 2.9.1 Non-compliant sampling for NACK signal from smartcard

#### Description

According to ISO/IEC 7816-3 standard, when a character parity error is detected, the receiver must assert a NACK signal, by pulling the transmit line low for one ETU period, at 10.3 to 10.7 ETU after the character START bit falling edge. The transmitter is expected to sample the line for NACK (for low level) from 10.8 to 11.2 ETU after the character START bit falling edge.

Instead, the USART peripheral in smartcard mode samples the transmit line for NACK from 10.3 to 10.7 ETU after the character START bit falling edge. This is unlikely to cause issues with receivers (smartcards) that respect the ISO/IEC 7816-3 standard. However, it may cause issues with respect to certification.

#### Workaround

None.

### 2.9.2 Break request preventing TC flag from being set

#### Description

After the end of transmission of data (D1), the transmission complete (TC) flag is not set when the following condition is met:

- CTS hardware flow control is enabled
- D1 transmission is in progress
- a break transfer is requested before the end of D1 transfer
- CTS is de-asserted before the end of D1 transfer

As a consequence, an application relying on the TC flag fails to detect the end of data transfer.

#### Workaround

In the application, only allow break request after the TC flag is set.

### 2.9.3 RTS is active while RE = 0 or UE = 0

#### Description

The RTS line is driven low as soon as RTSE bit is set, even if the USART is disabled (UE = 0) or the receiver is disabled (RE = 0), that is, not ready to receive data.

#### Workaround

Upon setting the UE and RE bits, configure the I/O used for RTS into alternate function.

### 2.9.4 Receiver timeout counter wrong start in two-stop-bit configuration

#### Description

In two-stop-bit configuration, the receiver timeout counter starts counting from the end of the second stop bit of the last character instead of starting from the end of the first stop bit.

#### Workaround

Subtract one bit duration from the value in the RTO bitfield of the USARTx\_RTOR register.

### 2.9.5 Data corruption due to noisy receive line

#### Description

In all modes, except synchronous slave mode, the received data may be corrupted if a glitch to zero shorter than the half-bit occurs on the receive line within the second half of the stop bit.

#### Workaround

Apply one of the following measures:

- Either use a noiseless receive line, or
- add a filter to remove the glitches if the receive line is noisy.

## 2.9.6 Received data may be corrupted upon clearing the ABREN bit

### Description

The USART receiver may miss data or receive corrupted data when the auto baud rate feature is disabled by software (ABREN bit cleared in the USART\_CR2 register) after an auto baud rate detection, while a reception is ongoing.

### Workaround

Do not clear the ABREN bit.

## 2.9.7 Noise error flag set while ONEBIT is set

### Description

When the ONEBIT bit is set in the USART\_CR3 register (one sample bit method is used), the noise error (NE) flag must remain cleared. Instead, this flag is set upon noise detection on the START bit.

### Workaround

None.

*Note: Having noise on the START bit is contradictory with the fact that the one sample bit method is used in a noise free environment.*

## 2.9.8 DMA channel 3 (CH3) not functional when USART2\_RX used for data reception

### Description

When USART2 uses DMA channel 3 for data reception, data transfers are blocked after the first byte has been received. As a result, DMA channel 3 cannot be used for USART2 data reception.

### Workaround

Use DMA channel 5 (CH5) for USART2 data reception.

## 2.10 LPUART

### 2.10.1 Break request preventing TC flag from being set

#### Description

After the end of transmission of data (D1), the transmission complete (TC) flag is not set when the following condition is met:

- CTS hardware flow control is enabled
- D1 transmission is in progress
- a break transfer is requested before the end of D1 transfer
- CTS is de-asserted before the end of D1 transfer

As a consequence, an application relying on the TC flag fails to detect the end of data transfer.

#### Workaround

In the application, only allow break request after the TC flag is set.

### 2.10.2 RTS is active while RE = 0 or UE = 0

#### Description

The RTS line is driven low as soon as RTSE bit is set, even if the LPUART is disabled (UE = 0) or the receiver is disabled (RE = 0), that is, not ready to receive data.

### Workaround

Upon setting the UE and RE bits, configure the I/O used for RTS into alternate function.

## 2.10.3 Receiver timeout counter wrong start in two-stop-bit configuration

### Description

In two-stop-bit configuration, the receiver timeout counter starts counting from the end of the second stop bit of the last character instead of starting from the end of the first stop bit.

### Workaround

Subtract one bit duration from the value in the RTO bitfield of the USARTx\_RTOR register.

## 2.10.4 Data corruption due to noisy receive line

### Description

In all modes, except synchronous slave mode, the received data may be corrupted if a glitch to zero shorter than the half-bit occurs on the receive line within the second half of the stop bit.

### Workaround

Apply one of the following measures:

- Either use a noiseless receive line, or
- add a filter to remove the glitches if the receive line is noisy.

## 2.10.5 Possible LPUART transmitter issue when using low BRR[15:0] value

### Description

The LPUART transmitter bit length sequence is not reset between consecutive bytes, which could result in a jitter that cannot be handled by the receiver device. As a result, depending on the receiver device bit sampling sequence, a desynchronization between the LPUART transmitter and the receiver device may occur resulting in data corruption on the receiver side.

This happens when the ratio between the LPUART kernel clock and the baud rate programmed in the LPUART\_BRR register (BRR[15:0]) is not an integer, and is in the three to four range. A typical example is when the 32.768 kHz clock is used as kernel clock and the baud rate is equal to 9600 baud, resulting in a ratio of 3.41.

### Workaround

Apply one of the following measures:

- On the transmitter side, increase the ratio between the LPUART kernel clock and the baud rate. To do so:
  - Increase the LPUART kernel clock frequency, or
  - Decrease the baud rate.
- On the receiver side, generate the baud rate by using a higher frequency and applying oversampling techniques if supported.

## 2.10.6 DMA channel 5 (CH5) not functional when LPUART1\_RX used for data reception

### Description

When LPUART1 uses DMA channel 5 for data reception, the data transfer is blocked after the first byte has been received. As a result, DMA channel 5 cannot be used for LPUART1 data reception.

### Workaround

Use DMA channel 3 (CH3) for LPUART1 data reception.

## 2.11 SPI/I2S

### 2.11.1 BSY bit may stay high when SPI is disabled

#### Description

The BSY flag may remain high upon disabling the SPI while operating in:

- master transmit mode and the TXE flag is low (data register full).
- master receive-only mode (simplex receive or half-duplex bidirectional receive phase) and an SCK strobing edge has not occurred since the transition of the RXNE flag from low to high.
- slave mode and NSS signal is removed during the communication.

#### Workaround

When the SPI operates in:

- master transmit mode, disable the SPI when TXE = 1 and BSY = 0.
- master receive-only mode, ignore the BSY flag.
- slave mode, do not remove the NSS signal during the communication.

### 2.11.2 BSY bit may stay high at the end of data transfer in slave mode

#### Description

BSY flag may sporadically remain high at the end of a data transfer in slave mode. This occurs upon coincidence of internal CPU clock and external SCK clock provided by master.

In such an event, if the software only relies on BSY flag to detect the end of SPI slave data transaction (for example to enter low-power mode or to change data line direction in half-duplex bidirectional mode), the detection fails.

As a conclusion, the BSY flag is unreliable for detecting the end of data transactions.

#### Workaround

Depending on SPI operating mode, use the following means for detecting the end of transaction:

- When NSS hardware management is applied and NSS signal is provided by master, use NSS flag.
- In SPI receiving mode, use the corresponding RXNE event flag.
- In SPI transmit-only mode, use the BSY flag in conjunction with a timeout expiry event. Set the timeout such as to exceed the expected duration of the last data frame and start it upon TXE event that occurs with the second bit of the last data frame. The end of the transaction corresponds to either the BSY flag becoming low or the timeout expiry, whichever happens first.

Prefer one of the first two measures to the third as they are simpler and less constraining.

Alternatively, apply the following sequence to ensure reliable operation of the BSY flag in SPI transmit mode:

1. Write last data to data register.
2. Poll the TXE flag until it becomes high, which occurs with the second bit of the data frame transfer.
3. Disable SPI by clearing the SPE bit mandatorily before the end of the frame transfer.
4. Poll the BSY bit until it becomes low, which signals the end of transfer.

**Note:** *The alternative method can only be used with relatively fast CPU speeds versus relatively slow SPI clocks or/and long last data frames. The faster is the software execution, the shorter can be the duration of the last data frame.*

### 2.11.3 Corrupted last bit of data and/or CRC, received in master mode with delayed SCK feedback

#### Description

In receive transaction, in both I<sup>2</sup>S and SPI master modes, the last bit of the transacted frame is not captured when signal provided by internal feedback loop from the SCK pin exceeds a critical delay. The lastly transacted bit of the stored data then keeps value from the previously received pattern. As a consequence, the last receive data bit may be wrong and/or the CRCERR flag can be unduly asserted in the SPI mode if any data under check sum and/or the CRC pattern is wrongly captured.

In SPI mode, data are synchronous with the APB clock. A delay of up to two APB clock periods can thus be tolerated for the internal feedback delay. The I<sup>2</sup>S mode is more sensitive than the SPI mode, and in particular when I2S prescaler is set to divide by an odd number and APB clock frequency is half the system clock frequency. In this case, margin of the internal feedback delay is lower than 1.5 APB clock periods.

Main factors contributing to the delay increase are low VDD level, high temperature, high SCK pin capacitive load and low SCK I/O output speed. The SPI communication speed has no impact.

The following table gives the maximum allowable APB frequency versus GPIOx\_OSPEEDR output speed bitfield setting for the SCK pin, at 30pF of capacitive load. The operation is safe up to that frequency.

**Table 5. Maximum allowable APB frequency at 30 pF load**

GPIOx_OSPEEDR[1:0] for SCK pin	Max. APB frequency for SPI mode [MHz]	Max. APB frequency for I <sup>2</sup> S mode [MHz]
11 (very high)	32	32
10 (high)	28 (32 at V <sub>DD</sub> > 1.8 V)	28 (32 at V <sub>DD</sub> > 1.8 V)
01 (medium)	12 (14 at V <sub>DD</sub> > 1.8 V)	8 (10 at V <sub>DD</sub> > 1.8 V)
00 (low)	1.5 (2 at V <sub>DD</sub> > 1.8 V)	1 (1.5 at V <sub>DD</sub> > 1.8 V)

#### Workaround

The following measures can be adopted, jointly or individually:

- Decrease the APB clock speed.
- Configure the I/O pad of the SCK pin to higher speed.

### 2.11.4 Corrupted last bit of data and/or CRC, received in master mode with delayed SCK feedback

#### Description

In master receive transaction, the last bit of the transacted frame is not captured when signal provided by internal feedback loop from the SCK pin exceeds a critical delay. The lastly transacted bit of the stored data then keeps value from the previously received pattern. As a consequence, the last receive data bit may be wrong and/or the CRCERR flag can be unduly asserted if any data under check sum and/or the CRC pattern is wrongly captured.

A delay of up to two APB clock periods can be tolerated for the internal feedback delay.

Main factors contributing to the delay increase are low VDD level, high temperature, high SCK pin capacitive load and low SCK I/O output speed. The SPI communication speed has no impact.

The following table gives the maximum allowable APB frequency versus GPIOx\_OSPEEDR output speed bitfield setting for the SCK pin, at 30pF of capacitive load. The operation is safe up to that frequency.

#### Workaround

The following measures can be adopted, jointly or individually:

- Decrease the APB clock speed.
- Configure the I/O pad of the SCK pin to higher speed.



## 2.11.5 SPI CRC corruption upon DMA transaction completion by another peripheral

### Description

When the following conditions are all met:

- CRC function for the SPI is enabled
- SPI transaction managed by software (as opposed to DMA) is ongoing and CRCNEXT flag set
- another peripheral using the DMA channel on which the SPI is mapped completes a DMA transfer, the CRCNEXT bit is unexpectedly cleared and the SPI CRC calculation may be corrupted, setting the CRC error flag.

### Workaround

Ensure that the DMA channel on which the SPI is mapped is not concurrently in use by another peripheral.

## 2.11.6 In I<sup>2</sup>S slave mode, enabling I2S while WS is active causes desynchronization

### Description

In I<sup>2</sup>S slave mode, the WS signal level is used to start the communication. If the I2S peripheral is enabled while the WS line is active (low for I<sup>2</sup>S protocol, high for LSB- or MSB-justified mode), and if the master is already sending the clock, the I2S peripheral (slave) starts communicating data from the instant of its enable, which causes desynchronization between the master and the slave throughout the whole communication.

### Workaround

Enable I2S peripheral while the WS line is at:

- high level, for I<sup>2</sup>S protocol.
- low level, for LSB- or MSB-justified mode.

## 2.11.7 Wrong CRC in full-duplex mode handled by DMA with imbalanced setting of data counters

### Description

When SPI is handled by DMA in full-duplex master or slave mode with CRC enabled, the CRC computation may temporarily freeze for the ongoing frame, which results in corrupted CRC.

This happens when the receive counter reaches zero upon the receipt of the CRC pattern (as the receive counter was set to a value greater, by CRC length, than the transmit counter). An internal signal dedicated to receive-only mode is left unduly pending. Consequently, the signal can cause the CRC computation to freeze during a next transaction in which DMA TXE event service is accidentally delayed (for example, due to DMA servicing a request from another channel).

### Workaround

Apply one of the following measures prior to each full-duplex SPI transaction:

- Set the DMA transmission and reception data counters to equal values. Upon the transaction completion, read the CRC pattern out from RxFIFO separately by software.
- Reset the SPI peripheral via peripheral reset register.

## 2.11.8 CRC error in SPI slave mode if internal NSS changes before CRC transfer

### Description

Some reference manual revisions may omit the information that the device operating as SPI slave must be configured in software NSS control if the SPI master pulses the NSS (for example in NSS pulse mode).

Otherwise, the transition of the internal NSS signal after the CRCNEXT flag is set might result in wrong CRC value computed by the device and, as a consequence, in a CRC error. As a consequence, the NSS pulse mode cannot be used along with the CRC function.

This is a documentation error rather than a product limitation.

## Workaround

No application workaround is required as long as the device operating as SPI slave is duly configured in software NSS control.

### 2.11.9 Anticipated communication upon SPI transit from slave receiver to master

#### Description

Regardless of the master mode configured, the communication clock starts upon setting the MSTR bit even though the SPI is disabled, if transiting from receive-only (RXONLY = 1) or half-duplex receive (BIDIMODE = 1 and BIDIOE = 0) slave mode to master mode.

#### Workaround

Apply one of the following measures:

- Before transiting to master mode, hardware-reset the SPI via the reset controller.
- Set the MSTR and SPE bits of the SPI configuration register simultaneously, which forces the immediate start of the communication clock. In transmitter configuration, load the data register in advance with the data to send.

## 2.12 USB

### 2.12.1 Possible packet memory overrun/underrun at low APB frequency

#### Description

Some data sheet and/or reference manual revisions may omit the information that 10 MHz minimum APB clock frequency is required to avoid USB data overrun/underrun issues.

Operating the USB peripheral with lower APB clock frequency may lead to:

- Overrun for out transactions - the USB peripheral fails to store the received data into the PBM before the next byte is received on the USB (PBM overrun). The USB cell detects an internal error condition, discards the last received byte, stops writing into the PBM, sends no acknowledge (forcing the host to retry the transaction), and informs the application by setting the PMAOVR flag/interrupt.
- Underrun for in transactions - the USB peripheral fails to read from the PBM the next byte to transmit before the transmission of the previous one is completed on the USB. The USB cell detects an internal error condition, stops reading from PBM, generates a bit stuffing error on the USB (forcing the host to retry the transaction), and informs the application by setting the PMAOVR flag/interrupt.

This is a documentation issue rather than a device limitation.

#### Workaround

No application workaround is required if the minimum APB clock frequency of 10 MHz is respected.

### 2.12.2 ESOF interrupt timing desynchronized after resume signaling

#### Description

Upon signaling resume, the device is expected to allow full 3 ms of time to the host or hub for sending the initial SOF (start of frame) packet, without triggering SUSP interrupt. However, the device only allows two full milliseconds and unduly triggers SUSP interrupt if it receives the initial packet within the third millisecond.

#### Workaround

When the device initiates resume (remote wake-up), mask the SUSP interrupt by setting the SUSPM bit for 3 ms, then unmask it by clearing SUSPM.

### 2.12.3 Incorrect CRC16 in the memory buffer

#### Description

Memory buffer locations are written starting from the address contained in the ADDRn\_RX for a number of bytes corresponding to the received data packet length, CRC16 inclusive (that is, data payload length plus two bytes), or up to the last allocated memory location defined by BL\_SIZE and NUM\_BLOCK, whichever comes first. In the former case, the CRC16 checksum is written wrongly, with its least significant byte going to both memory buffer byte locations expected to receive the least and the most significant bytes of the checksum.

Although the checksum written in the memory buffer is wrong, the underlying CRC checking mechanism in the USB peripheral is fully functional.

#### Workaround

Ignore the CRC16 data in the memory buffer.

### 2.12.4 Buffer description table update completes after CTR interrupt triggers

#### Description

During OUT transfers, the correct transfer interrupt (CTR) is triggered a little before the last USB SRAM accesses have completed. If the software responds quickly to the interrupt, the full buffer contents may not be correct.

#### Workaround

Software should ensure that a small delay is included before accessing the SRAM contents. This delay should be 800 ns in Full Speed mode and 6.4 µs in Low Speed mode.

### 2.12.5 USB BCD functionality limited below –20 °C

#### Description

Primary and secondary detection can return an incorrectly detected port type.

This limitation may be observed on a small number of devices when the temperature is below –20 °C.

#### Workaround

None.

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## Revision history

**Table 6. Document revision history**

Date	Version	Changes
03-Jun-2019	1	Initial release.
15-Dec-2021	2	Distinction made between device and documentation errata. Added DMA, TIM, LPTIM, and USB errata. Updated System, ADC, RTC and TAMP, I2C, SPI/I2S, USART, and LPUART errata.
12-Aug-2024	3	Added errata: <ul style="list-style-type: none"> <li>Wake-up sequence from Standby mode fails when using more than one wake-up source</li> <li>LSE crystal oscillator may be disturbed by transitions on PC13</li> <li>GPIO locking mechanism failed if critical address is read</li> <li>Writing ADC_CFGR2 register while ADEN bit is set resets CKMODE[1:0] bitfield</li> <li>A tamper event fails to trigger timestamp or timestamp overflow events during a few cycles after clearing TSF</li> <li>SDA held low upon SMBus timeout expiry in slave mode</li> <li>Received data may be corrupted upon clearing the ABREN bit</li> <li>Noise error flag set while ONEBIT is set</li> <li>Possible LPUART transmitter issue when using low BRR[15:0] value</li> <li>Buffer description table update completes after CTR interrupt triggers</li> </ul>

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