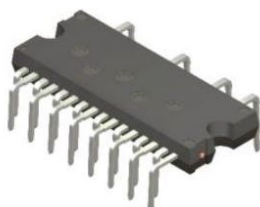


SLLIMM-nano 2nd series IPM, 3-phase inverter, 3 A, 1.6 Ω max., 600 V, N-channel MDmesh DM2 Power MOSFET



N2DIP-26L
type Z no stand-off



Product status link

STIPQ3M60T-HZS

Product summary

Order code	STIPQ3M60T-HZS
Marking	IPQ3M60T-HZS
Package	N2DIP-26L type Z no stand-off
Packing	Tube

Features

- IPM 3 A, 600 V, $R_{DS(on)} = 1.6 \Omega$, 3-phase Power MOSFET inverter bridge including control ICs for gate driving
- Optimized for low electromagnetic interference
- 3.3 V, 5 V, 15 V CMOS/TTL input comparators with hysteresis and pull-down/pull-up resistors
- Undervoltage lockout
- Internal bootstrap diode
- Interlocking function
- Shutdown function
- Comparator for fault protection against overtemperature and overcurrent
- Op-amp for advanced current sensing
- Optimized pinout for easy board layout
- NTC for temperature control (UL 1434 CA 2 and 4)
- Isolation ratings of 1800 Vrms/min.
- UL recognition: UL 1557 file, E81734

Applications

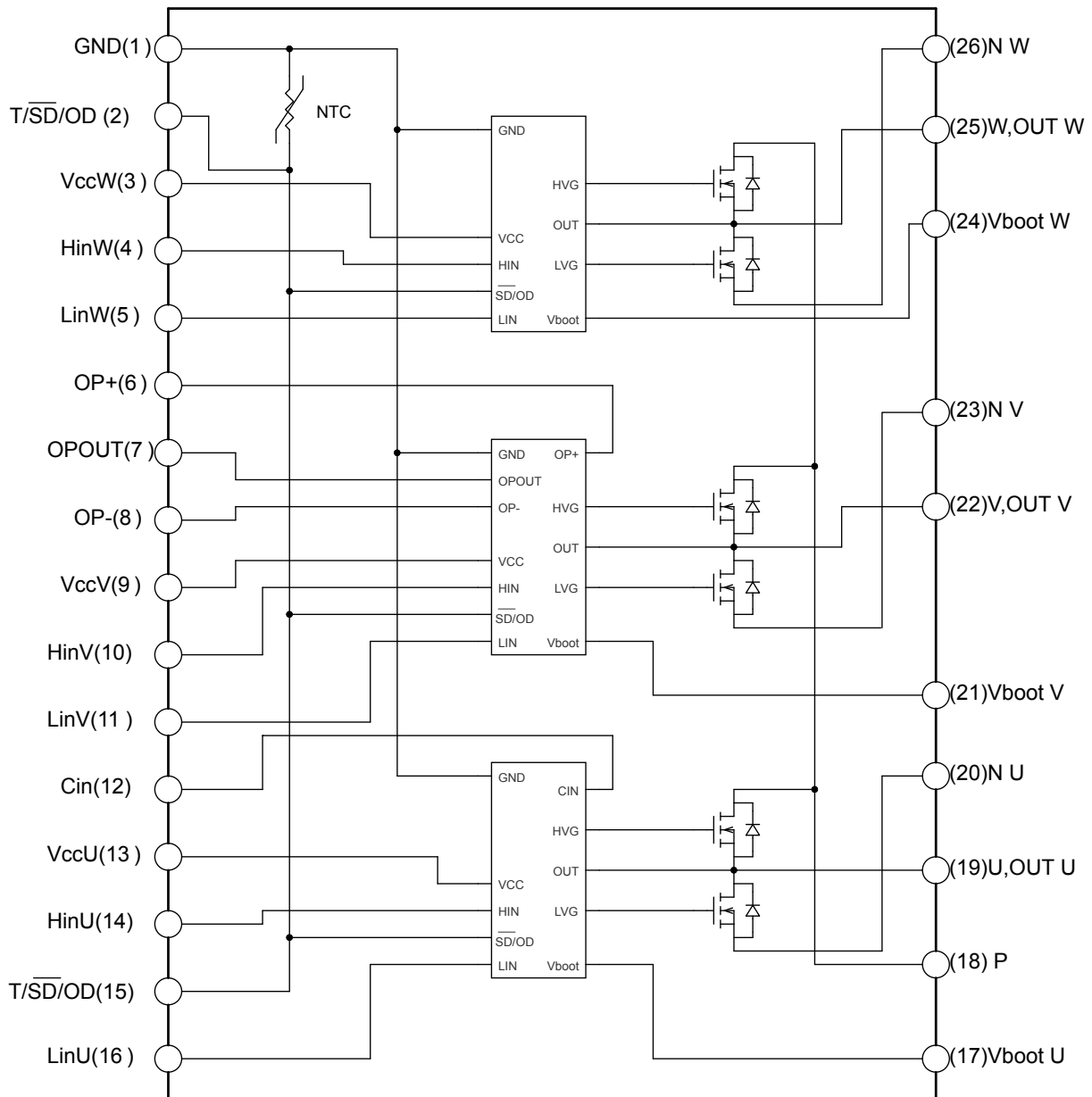
- 3-phase induction motor (ACIM)
- Dishwasher
- Fans
- Kitchen hoods
- PMSM / BLDC motor control
- Refrigerators and freezers

Description

This SLLIMM (small low-loss intelligent molded module)-nano provides a compact, high-performance AC motor drive in a simple, rugged design. It is composed of six N-channel MDmesh DM2 Power MOSFETs with intrinsic fast-recovery diode and three half-bridge HVICs for gate driving, providing low electromagnetic interference (EMI) characteristics with optimized switching speed. The package is designed to allow a better and easy screw on heatsink. It is optimized for thermal performance and compactness in built-in motor applications, or other low-power applications where assembly space is limited. This IPM includes an operational amplifier, completely uncommitted, and a comparator that can be used to design a fast and efficient protection circuit.

1 Internal schematic diagram and pin configuration

Figure 1. Internal schematic diagram

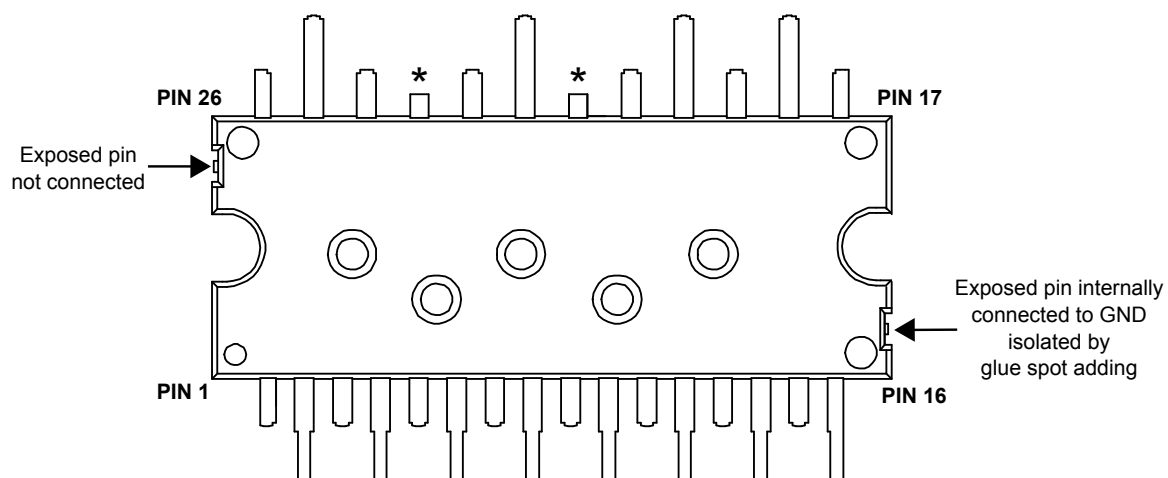


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Table 1. Pin description

Pin	Symbol	Description
1	GND	Ground
2	T/ $\overline{\text{SD}}$ /OD	NTC thermistor terminal/shutdown logic input (active low)/open-drain (comparator output)
3	V _{CC} W	Low-voltage power supply W phase
4	HIN W	High-side logic input for W phase
5	LIN W	Low-side logic input for W phase
6	OP+	Op-amp non-inverting input
7	OP _{OUT}	Op-amp output
8	OP-	Op-amp inverting input
9	V _{CC} V	Low-voltage power supply V phase
10	HIN V	High-side logic input for V phase
11	LIN V	Low-side logic input for V phase
12	CIN	Comparator input
13	V _{CC} U	Low-voltage power supply for V phase
14	HIN U	High-side logic input for V phase
15	T/ $\overline{\text{SD}}$ /OD	NTC thermistor terminal/shutdown logic input (active low)/open-drain (comparator output)
16	LIN U	Low-side logic input for U phase
17	V _{boot} U	Bootstrap voltage for U phase
18	P	Positive DC input
19	U, OUT _U	U phase output
20	N _U	Negative DC input for U phase
21	V _{boot} V	Bootstrap voltage for V phase
22	V, OUT _V	V phase output
23	N _V	Negative DC input for V phase
24	V _{boot} W	Bootstrap voltage for W phase
25	W, OUT _W	W phase output
26	N _W	Negative DC input for W phase

Figure 2. Pin layout (top view) - N2DIP-26L type Z



* Dummy pins internally connected to P (positive DC input)

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2 Electrical ratings

$T_J = 25\text{ °C}$ unless otherwise specified

2.1 Absolute maximum ratings

Table 2. Inverter part

Symbol	Parameter	Value	Unit
V_{DSS}	MOSFET blocking voltage (or drain-source voltage) for each MOSFET ($V_{IN}^{(1)} = 0$)	600	V
$\pm I_D$	Continuous drain current for each MOSFET ($T_C = 25\text{ °C}$)	3	A
$\pm I_{DP}^{(2)}$	Peak drain current for each MOSFET (less than 1 ms)	6	A
P_{TOT}	Total power dissipation for each MOSFET ($T_C = 25\text{ °C}$)	12.5	W

1. Applied among $HINx$, $LINx$ and GND for $x = U, V, W$
2. Pulse width limited by maximum junction temperature

Table 3. Control part

Symbol	Parameter	Min.	Max.	Unit
V_{CC}	Low voltage power supply	-0.3	21	V
V_{boot}	Bootstrap voltage	-0.3	620	V
V_{OUT}	Output voltage applied among OUT_U , OUT_V , OUT_W - GND	$V_{boot} - 21$	$V_{boot} + 0.3$	V
V_{CIN}	Comparator input voltage	-0.3	$V_{CC} + 0.3$	V
V_{op+}	Op-amp non-inverting input	-0.3	$V_{CC} + 0.3$	V
V_{op-}	Op-amp inverting input	-0.3	$V_{CC} + 0.3$	V
V_{IN}	Logic input voltage applied among $HINx$, $LINx$ and GND	-0.3	15	V
$V_{T/SD/OD}$	Open-drain voltage	-0.3	15	V
dV_{out}/dt	Allowed output slew rate		50	V/ns

Table 4. Total system

Symbol	Parameter	Value	Unit
V_{ISO}	Isolation withstand voltage applied on each pin and heat sink plate (AC voltage, $t = 60\text{ s}$)	1800	Vrms
T_J	Power chip operating junction temperature range	-40 to 150	°C
T_C	Module case operating temperature range	-40 to 125	°C

2.2 Thermal data

Table 5. Thermal data

Symbol	Parameter	Value	Unit
R_{thJC}	Thermal resistance, junction-to-case, single MOSFET	10	°C/W

3 Electrical characteristics

$T_J = 25\text{ }^{\circ}\text{C}$ unless otherwise specified.

3.1 Inverter part

Table 6. Static

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{DSS}	Zero-gate voltage drain current	$V_{DS} = 600\text{ V}$, $V_{CC} = 15\text{ V}$, $V_{boot} = 15\text{ V}$			1	mA
$V_{(BR)DSS}$	Drain-source breakdown voltage	$V_{CC} = V_{boot} = 15\text{ V}$, $V_{IN}^{(1)} = 0\text{ V}$, $I_D = 1\text{ mA}$	600			V
$R_{DS(on)}$	Static drain source turn-on resistance	$V_{CC} = V_{boot} = 15\text{ V}$, $V_{IN}^{(1)} = 0\text{ to }5\text{ V}$, $I_D = 1.5\text{ A}$		1.35	1.6	Ω
V_{SD}	Drain-source diode forward voltage	$V_{IN}^{(1)} = 0$ "logic state", $I_D = 3\text{ A}$		1.4	1.9	V

1. Applied among $HINx$, $LINx$ and GND for $x = U, V, W$.

Table 7. Inductive load switching time and energy

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{on}^{(1)}$	Turn-on time	$V_{DD} = 300\text{ V}$, $V_{CC} = V_{boot} = 15\text{ V}$, $V_{IN}^{(2)} = 0\text{ to }5\text{ V}$, $I_C = 1.5\text{ A}$ (see Figure 4. Switching time definition)	-	220	-	ns
$t_{c(on)}^{(1)}$	Crossover time (on)		-	72	-	
$t_{off}^{(1)}$	Turn-off time		-	225	-	
$t_{c(off)}^{(1)}$	Crossover time (off)		-	29	-	
t_{rr}	Reverse recovery time		-	79	-	
E_{on}	Turn-on switching energy		-	47	-	μJ
E_{off}	Turn-off switching energy		-	3.9	-	

1. t_{ON} and t_{OFF} include the propagation delay time of the internal drive. $t_{C(ON)}$ and $t_{C(OFF)}$ are the switching times of the MOSFET itself under the internally given gate driving conditions.

2. Applied among $HINx$, $LINx$ and GND for $x = U, V, W$.

The diagram illustrates a buck converter circuit using the GIPD161120151702R IC. The IC is a 16-pin package with the following pins and connections:

- LIN**: Connected to a 5V square wave input.
- Vboot**: Connected to a bootstrap capacitor, which is also connected to the **SD/OD** pin. The bootstrap capacitor is charged by a 5V regulator.
- SD/OD**: Connected to a 5V regulator.
- HIN**: Connected to the gate of the high-side MOSFET.
- VCC**: Connected to a 5V regulator.
- OUT**: Connected to the drain of the high-side MOSFET and the inductor **L**.
- LVG**: Connected to the gate of the low-side MOSFET.
- GND**: Connected to ground.

The output of the converter is a square wave with a peak current I_O . The output voltage is regulated by a feedback loop. The output capacitor **C** is connected to the output of the converter and ground. The output voltage is V_{dd} .

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3.2 Control part

Table 8. Low-voltage power supply ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC_hys}	V_{CC} UV hysteresis		1.2	1.5	1.8	V
V_{CC_thON}	V_{CC} UV turn-ON threshold		11.5	12	12.5	V
V_{CC_thOFF}	V_{CC} UV turn-OFF threshold		10	10.5	11	V
I_{qccu}	Undervoltage quiescent supply current	$V_{CC} = 10\text{ V}$, $T/\overline{SD}/OD = 5\text{ V}$, $LIN = HIN = CIN = 0\text{ V}$			150	μA
I_{qcc}	Quiescent current	$V_{CC} = 10\text{ V}$, $T/\overline{SD}/OD = 5\text{ V}$, $LIN = HIN = CIN = 0\text{ V}$			1	mA
V_{ref}	Internal comparator (CIN) reference voltage		0.51	0.54	0.56	V

Table 9. Bootstrapped voltage ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{BS_hys}	V_{BS} UV hysteresis		1.2	1.5	1.8	V
V_{BS_thON}	V_{BS} UV turn-ON threshold		11.1	11.5	12.1	V
V_{BS_thOFF}	V_{BS} UV turn-OFF threshold		9.8	10	10.6	V
I_{QBSU}	Undervoltage V_{BS} quiescent current	$V_{BS} < 9\text{ V}$, $T/\overline{SD}/OD = 5\text{ V}$, $LIN = 0\text{ V}$ and $HIN = 5\text{ V}$, $CIN = 0$		70	110	μA
I_{QBS}	V_{BS} quiescent current	$V_{BS} = 15\text{ V}$, $T/\overline{SD}/OD = 5\text{ V}$, $LIN = 0\text{ V}$ and $HIN = 5\text{ V}$, $CIN = 0$		200	300	μA
$R_{DS(on)}$	Bootstrap driver on-resistance	LVG ON		120		Ω

Table 10. Logic inputs ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{il}	Low logic level voltage				0.8	V
V_{ih}	High logic level voltage		2.25			V
I_{HINH}	HIN logic "1" input bias current	$HIN = 15\text{ V}$	20	40	100	μA
I_{HINI}	HIN logic "0" input bias current	$HIN = 0\text{ V}$			1	μA
I_{LINI}	LIN logic "1" input bias current	$LIN = 15\text{ V}$	20	40	100	μA
I_{LINh}	LIN logic "0" input bias current	$LIN = 0\text{ V}$			1	μA
I_{SDh}	$\overline{SD}$ logic "0" input bias current	$\overline{SD} = 15\text{ V}$	210	350	477	μA
I_{SDI}	$\overline{SD}$ logic "1" input bias current	$\overline{SD} = 0\text{ V}$			3	μA
Dt	Dead time	See Section 3.3 Waveform definitions		180		ns

Table 11. Op-amp characteristics ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{io}	Input offset voltage	$V_{ic} = 0\text{ V}$, $V_o = 7.5\text{ V}$			6	mV
I_{io}	Input offset current	$V_{ic} = 0\text{ V}$, $V_o = 7.5\text{ V}$		4	40	nA
I_{ib}	Input bias current ⁽¹⁾			100	200	nA
V_{OL}	Low-level output voltage	$R_L = 10\text{ k}\Omega$ to V_{CC}		75	150	mV
V_{OH}	High-level output voltage	$R_L = 10\text{ k}\Omega$ to GND	14	14.7		V
I_o	Output short-circuit current	Source, $V_{id} = +1\text{ V}$, $V_o = 0\text{ V}$	16	30		mA
		Sink, $V_{id} = -1\text{ V}$, $V_o = V_{CC}$	50	80		mA
SR	Slew rate	$V_i = 1$ to 4 V , $C_L = 100\text{ pF}$, unity gain	2.5	3.8		V/ μs
GBWP	Gain bandwidth product	$V_o = 7.5\text{ V}$	8	12		MHz
A_{vd}	Large signal voltage gain	$R_L = 2\text{ k}\Omega$	70	85		dB
SVR	Supply voltage rejection ratio	vs. V_{CC}	60	75		dB
CMRR	Common mode rejection ratio		55	70		dB

1. The direction of the input current is out of the IC.

Table 12. Sense comparator characteristics ($V_{CC} = 15\text{ V}$ unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{ib}	Input bias current	$V_{CIN} = 1\text{ V}$			1	μA
V_{od}	Open-drain low level output voltage	$I_{od} = 3\text{ mA}$			0.5	V
R_{ON_OD}	Open-drain low level output resistance	$I_{od} = 3\text{ mA}$		166		Ω
R_{PD_SD}	$\overline{SD}$ pull-down resistor ⁽¹⁾			125		k Ω
t_{d_comp}	Comparator delay	$T/\overline{SD}/OD$ pulled to 5 V through $100\text{ k}\Omega$ resistor		90	130	ns
SR	Slew rate	$C_L = 180\text{ pF}$, $R_{pu} = 5\text{ k}\Omega$		60		V/ μs
t_{sd}	Shutdown to high-/low-side driver propagation delay	$V_{OUT} = 0$, $V_{boot} = V_{CC}$, $V_{IN} = 0$ to 3.3 V	50	125	200	ns
t_{isd}	Comparator triggering to high-/low-side driver turn-off propagation delay	Measured applying a voltage step from 0 V to 3.3 V to pin CIN	50	200	250	

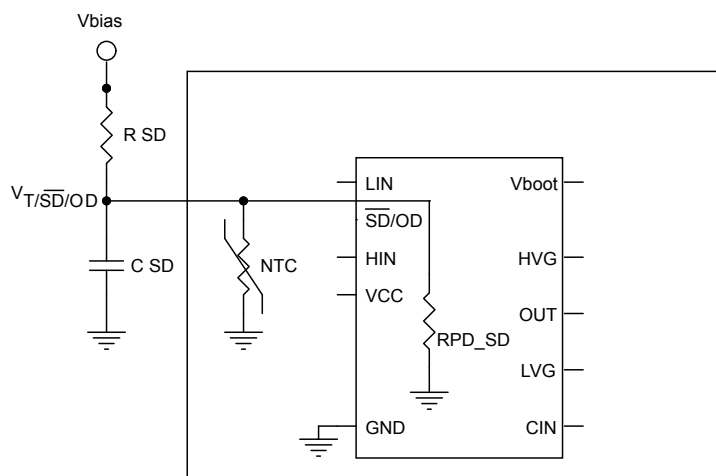
1. Equivalent values are the result of the resistances of three drivers in parallel.

Table 13. Truth table

Conditions	Logic input (V_I)			Output	
	$\overline{T/SD/OD}$	LIN	HIN	LVG	HVG
Shutdown enable half-bridge tri-state	L	X ⁽¹⁾	X ⁽¹⁾	L	L
Interlocking half-bridge tri-state	H	H	H	L	L
0 "logic state" half-bridge tri-state	H	L	L	L	L
1 "logic state" low-side direct driving	H	H	L	H	L
1 "logic state" high-side direct driving	H	L	H	L	H

1. X: do not care.

3.2.1 NTC thermistor

Figure 5. Internal structure of $\overline{SD}$ and NTC


RPD_SD: equivalent value as result of resistances of three drivers in parallel.

Figure 6. Equivalent resistance (NTC//R_{PD_SD})

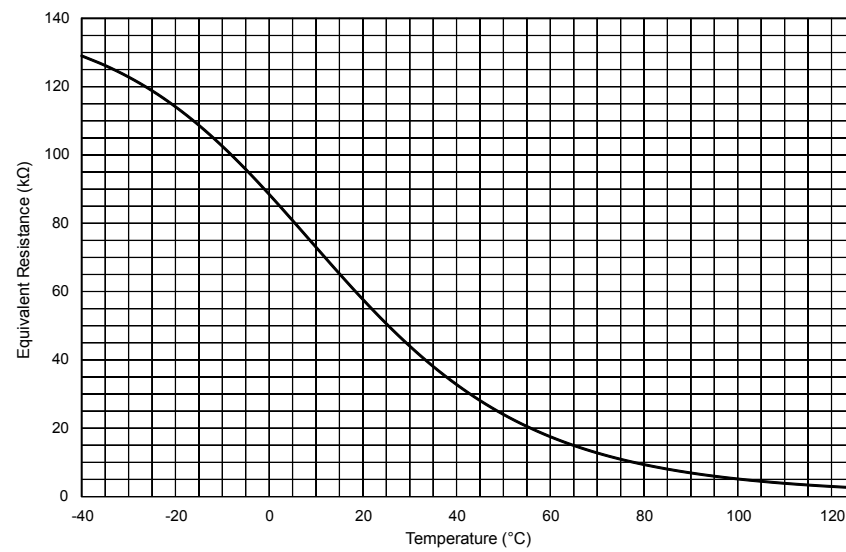


Figure 7. Equivalent resistance (NTC//R_{PD_SD}) zoom

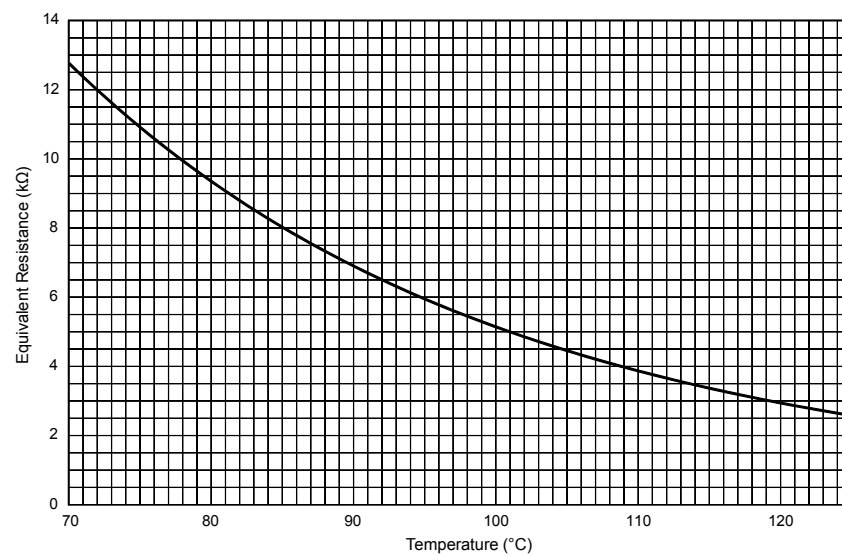
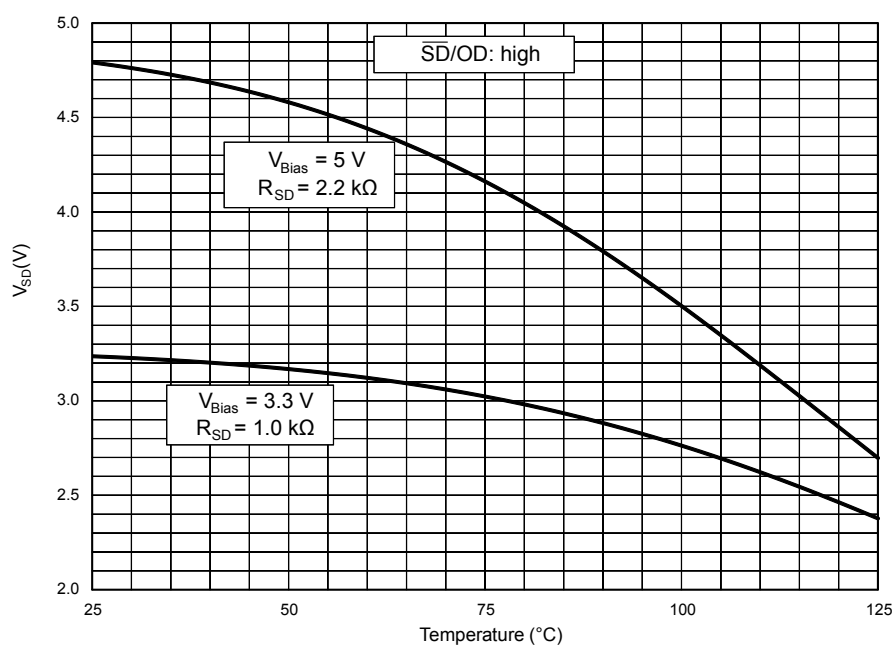
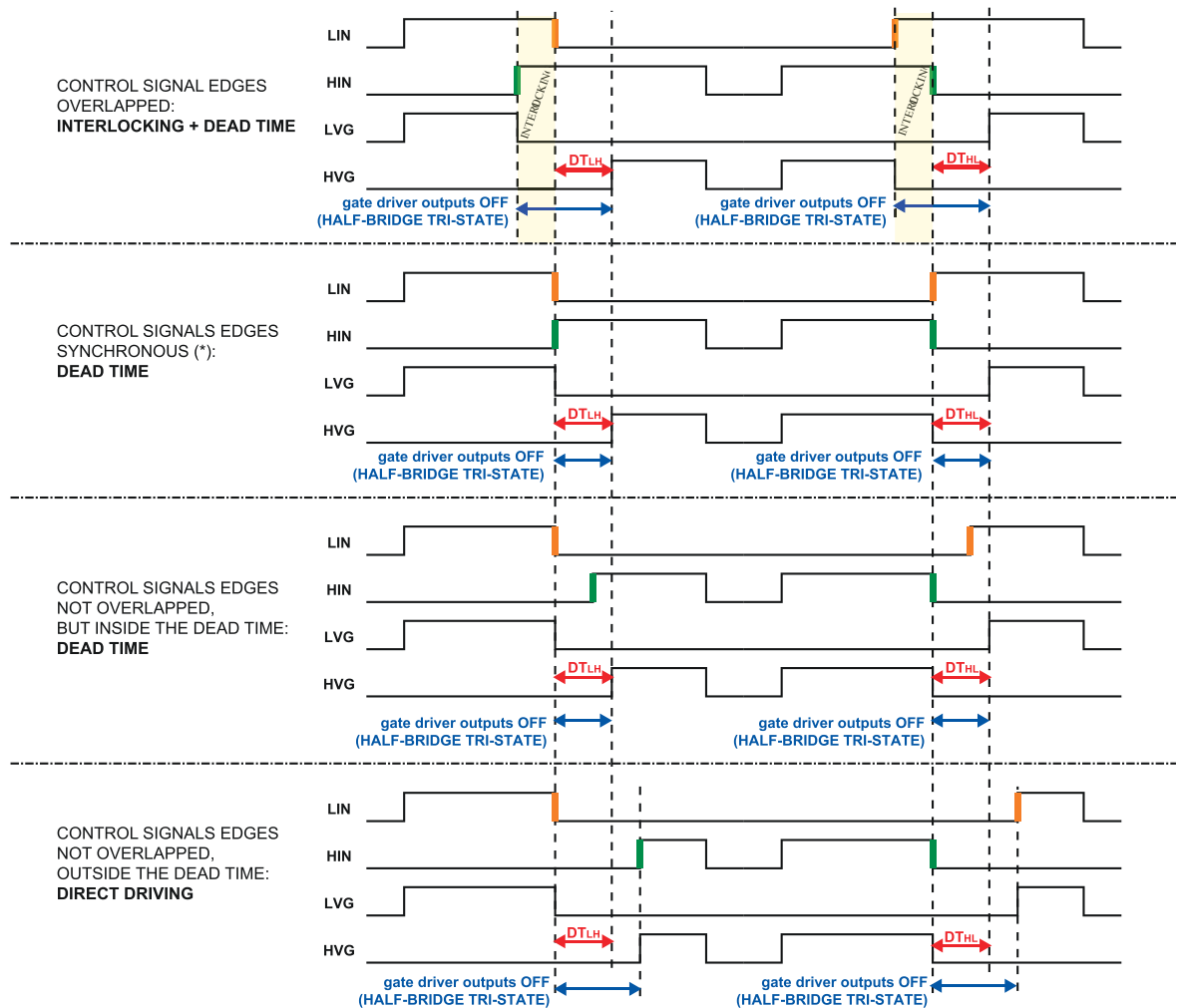


Figure 8. Voltage of T/SD/OD pin according to NTC temperature



3.3 Waveform definitions

Figure 9. Dead time and interlocking waveform definitions



4 Shutdown function

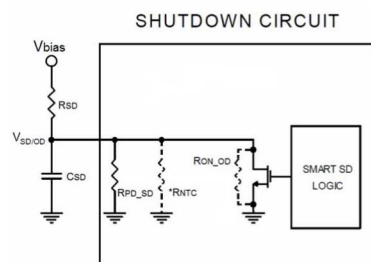
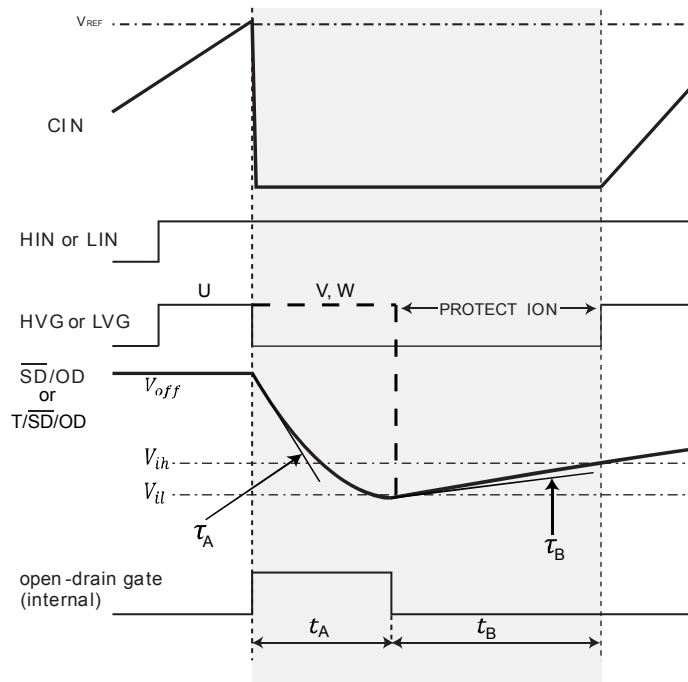
The device is equipped with three half-bridge IC gate drivers and integrates a comparator for fault detection. The comparator has an internal voltage reference V_{REF} connected to the inverting input, while the non-inverting input pin (CIN) can be connected to an external shunt resistor for current monitoring.

Since the comparator is embedded in the U IC gate driver, in case of fault it disables directly the U outputs, whereas the shutdown of V and W IC gate drivers depends on the RC value of the external SD circuitry, which fixes the disabling time.

For an effective design of the shutdown circuit, please refer to Application note AN4966.

Figure 10. Shutdown timing waveforms

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$$t_A \cong \tau_A \cdot \ln \left(\frac{V_{off} - V_{on}}{V_{il} - V_{on}} \right), \quad t_B \cong \tau_B \cdot \ln \left(\frac{V_{il} - V_{off}}{V_{ih} - V_{off}} \right)$$

$$\tau_A = (R_{ON_OD} // R_{SD} // R_{PD_SD} // R_{NTC}) \cdot C_{SD} \cong R_{ON_OD} \cdot C_{SD}$$

$$\tau_B = (R_{SD} // R_{PD_SD} // R_{NTC}) \cdot C_{SD}$$

$$V_{on} = \frac{R_{ON_OD} // R_{PD_SD} // R_{NTC}}{(R_{ON_OD} // R_{PD_SD} // R_{NTC}) + R_{SD}} \cdot V_{bias}$$

$$\cong \frac{R_{ON_OD}}{R_{ON_OD} + R_{SD}} \cdot V_{bias}$$

$$V_{off} = \frac{R_{PD_SD} // R_{NTC}}{(R_{PD_SD} // R_{NTC}) + R_{SD}} \cdot V_{bias}$$

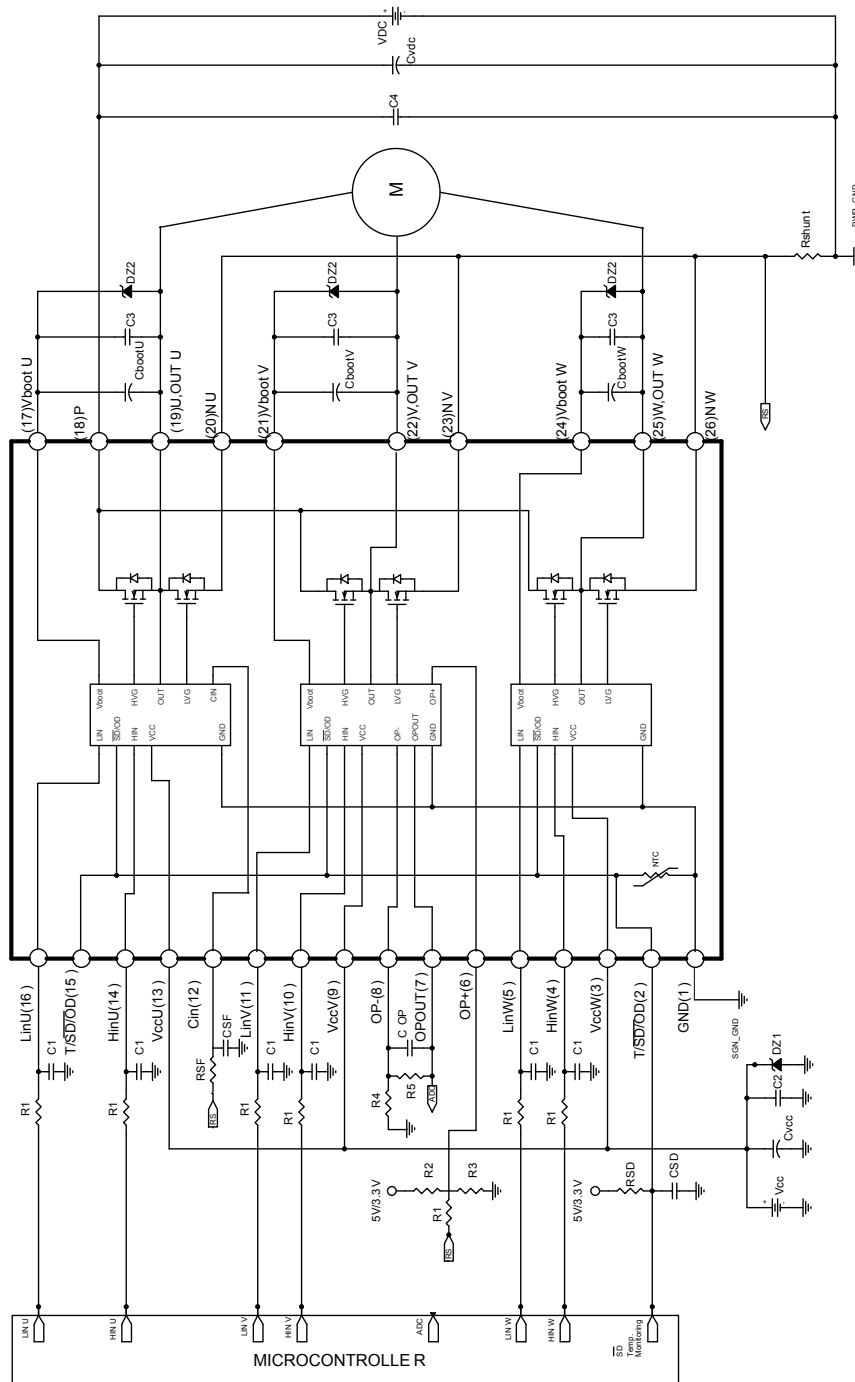
R_{SD} and C_{SD} external circuitry must be designed to ensure $V_{on} < V_{il}$ & $V_{off} > V_{ih}$

Please refer to AN4966 for further details.

* R_{NTC} to be considered only when the NTC is internally connected to the T/SD/OD pin.

5 Application circuit example

Figure 11. Application circuit example



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Application designers are free to use a different scheme according to the specifications of the device.

5.1 Guidelines

- Input signals HIN, LIN are active high logic. A 375 k Ω (typ.) pull-down resistor is built-in for each input. To avoid input signal oscillation, the wiring of each input should be as short as possible, and the use of RC filters (R_1 , C_1) on each input signal is suggested. The filters should be with a time constant of about 100 ns and placed as close as possible to the IPM input pins.
- The use of a bypass capacitor C_{VCC} (aluminum or tantalum) can reduce the transient circuit demand on the power supply. Also, to reduce any high-frequency switching noise distributed on the power lines, a decoupling capacitor C_2 (100 to 220 nF, with low ESR and low ESL) should be placed as close as possible to the V_{CC} pin and in parallel with the bypass capacitor.
- The use of an RC filter (R_{SF} , C_{SF}) is recommended to prevent protection circuit malfunction. The time constant ($R_{SF} \times C_{SF}$) should be set to 1 μ s and the filter must be placed as close as possible to the C_{IN} pin.
- The $\overline{SD}$ is an input/output pin (open-drain type if it is used as output). A built-in thermistor NTC is internally connected between the $\overline{SD}$ pin and GND. The voltage V_{SD-GND} decreases as the temperature increases, due to the pull-up resistor R_{SD} . In order to keep the voltage always higher than the high-level logic threshold, the pull-up resistor should be set to 1 k Ω or 2.2 k Ω for 3.3 V or 5 V MCU power supply, respectively. The capacitor C_{SD} of the filter on $\overline{SD}$ should be fixed no higher than 3.3 nF in order to assure the $\overline{SD}$ activation time $\tau_A \leq 500$ ns. Besides, the filter should be placed as close as possible to the $\overline{SD}$ pin.
- The decoupling capacitor C_3 (from 100 to 220 nF, ceramic with low ESR and low ESL), in parallel with each C_{boot} , filters high-frequency disturbance. Both C_{boot} and C_3 (if present) should be placed as close as possible to the U, V, W and V_{boot} pins. Bootstrap negative electrodes should be connected to U, V, W terminals directly and separated from the main output wires.
- To avoid overvoltage on the V_{CC} pin, a Zener diode (Dz1) can be used. Similarly on the V_{boot} pin, a Zener diode (Dz2) can be placed in parallel with each C_{boot} .
- The use of the decoupling capacitor C_4 (100 to 220 nF, with low ESR and low ESL) in parallel with the electrolytic capacitor C_{Vdc} is useful to prevent surge destruction. Both capacitors C_4 and C_{Vdc} should be placed as close as possible to the IPM (C_4 has priority over C_{Vdc}).
- By integrating an application-specific type HVIC inside the module, direct coupling to the MCU terminals without an opto-couplers is possible.
- Low-inductance shunt resistors have to be used for phase leg current sensing.
- In order to avoid malfunctions, the wiring on N pins, the shunt resistor and P_{WR_GND} should be as short as possible.
- The connection of SGN_GND to PWR_GND on one point only (close to the shunt resistor terminal) can reduce the impact of power ground fluctuation.

These guidelines ensure the specifications of the device for application designs. For further details, please refer to the relevant application note.

Table 14. Recommended operating conditions

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{PN}	Supply voltage	Applied among P-Nu, Nv, Nw		300	500	V
V_{CC}	Control supply voltage	Applied to V_{CC-GND}	13.5	15	18	V
V_{BS}	High-side bias voltage	Applied to $V_{BOOTx-OUT}$ for $x = U, V, W$	13		18	V
t_{dead}	Blanking time to prevent arm-short	For each input signal	1			μ s
f_{PWM}	PWM input signal	-40 $^{\circ}$ C < T_C < 100 $^{\circ}$ C -40 $^{\circ}$ C < T_J < 125 $^{\circ}$ C			25	kHz
T_C	Operating case temperature				100	$^{\circ}$ C

6 Electrical characteristics (curves)

Figure 12. Output characteristics

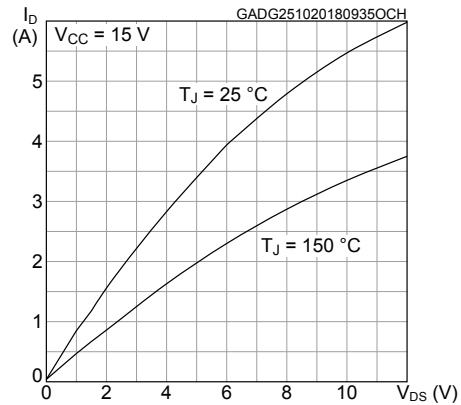


Figure 13. Diode V_F vs forward current

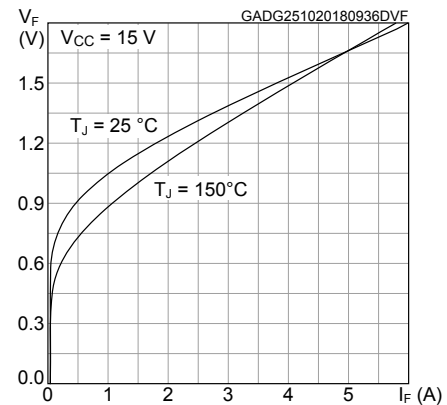


Figure 14. I_D vs case temperature

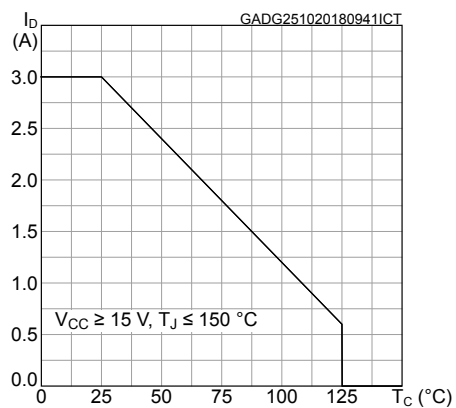


Figure 15. E_{ON} switching energy vs drain current

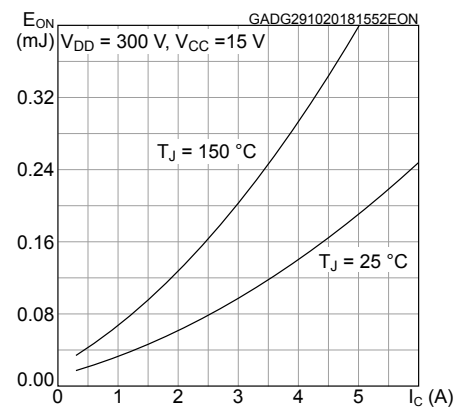


Figure 16. E_{OFF} switching energy vs drain current

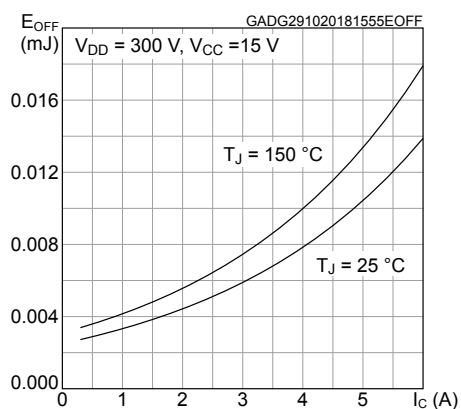
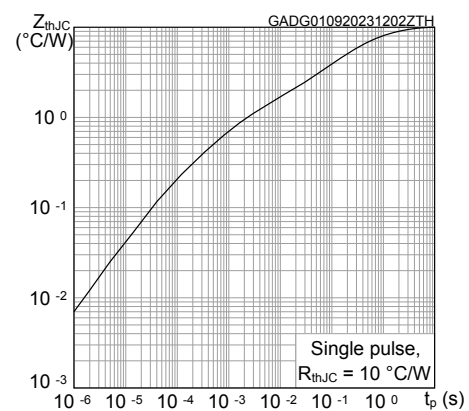


Figure 17. Maximum transient thermal impedance for MOSFET

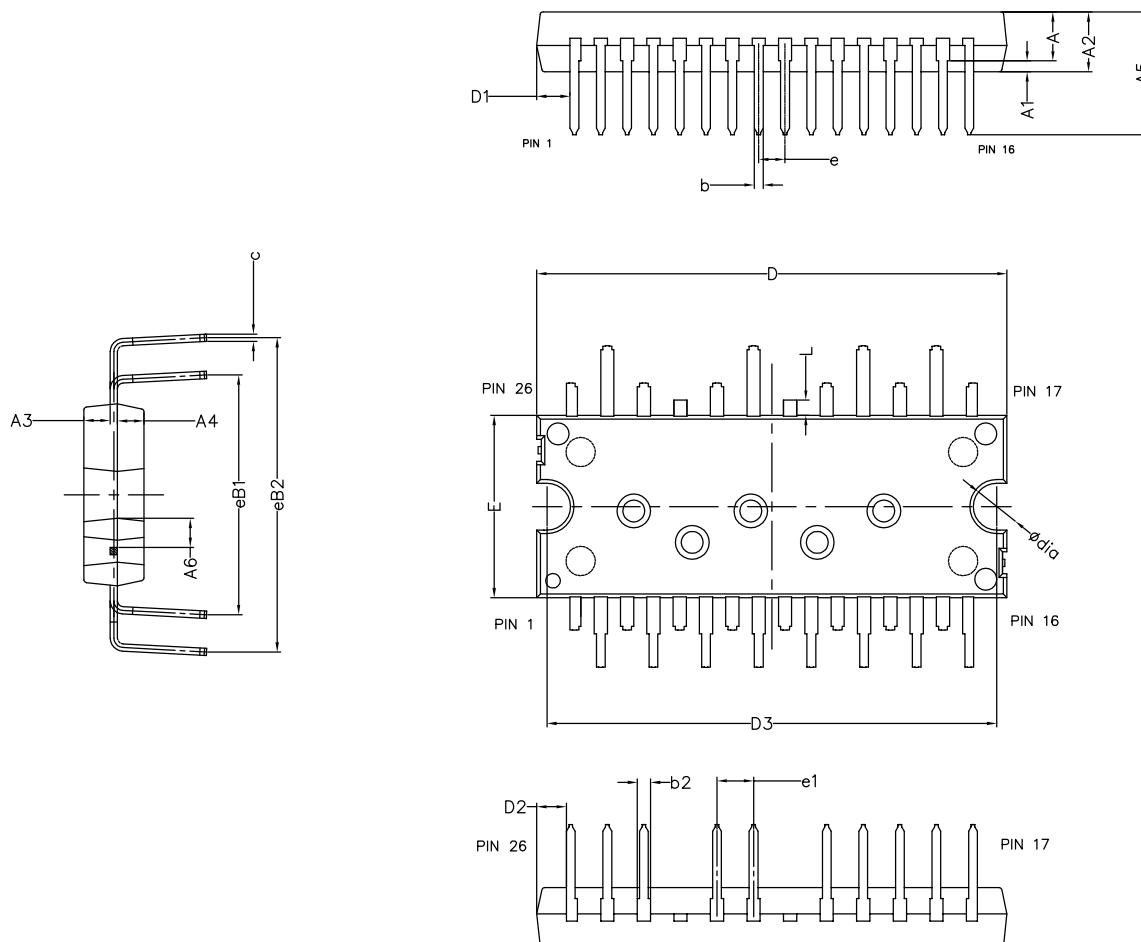


7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: www.st.com. ECOPACK is an ST trademark.

7.1 N2DIP-26L type Z no stand-off package information

Figure 18. N2DIP-26L type Z no stand-off package outline



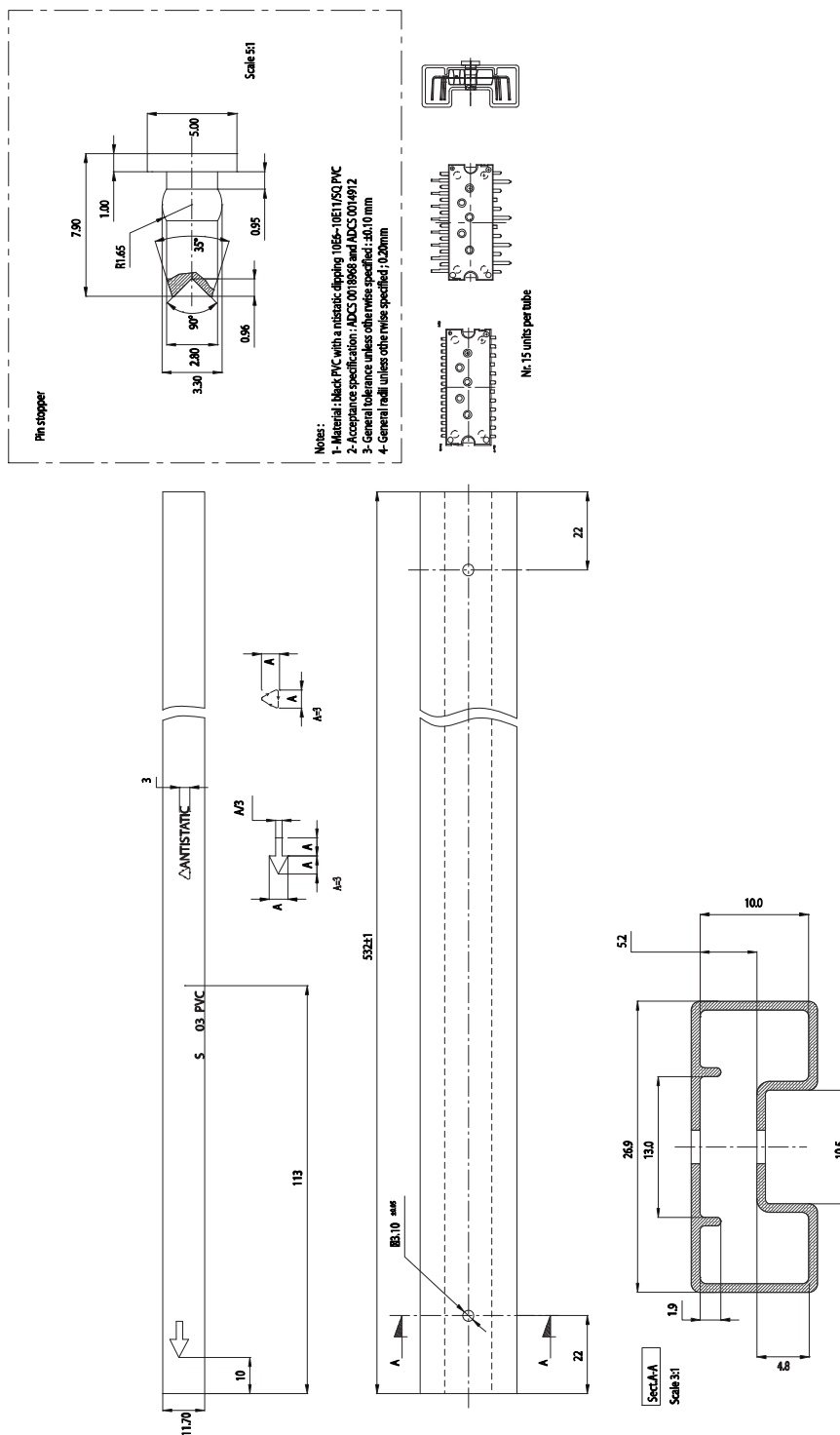
8558322_3_typeZ_NO_stand_off

Table 15. N2DIP-26L type Z no stand-off mechanical data

Dim.	mm		
	Min.	Typ.	Max.
A			3.80
A1	0.45	0.75	1.05
A2	4.00	4.10	4.20
A3	1.70	1.80	1.90
A4	1.70	1.80	1.90
A5	8.10	8.40	8.70
A6	1.75		
b	0.53		0.72
b2	0.83		1.02
c	0.46		0.59
D	32.05	32.15	32.25
D1	2.10		
D2	1.85		
D3	30.65	30.75	30.85
E	12.35	12.45	12.55
e	1.70	1.80	1.90
e1	2.40	2.50	2.60
eB1	16.10	16.40	16.70
eB2	21.18	21.48	21.78
L	0.85	1.05	1.25
Dia	3.10	3.20	3.30

7.2 N2DIP-26L packing information

Figure 19. N2DIP-26L tube (dimensions are in mm)



Revision history

Table 16. Document revision history

Date	Revision	Changes
03-May-2021	1	Initial release.
01-Sep-2023	2	Updated Features on cover page. Updated Table 2 . Inverter part, Table 4 . Total system, Table 5 . Thermal data and Figure 17 . Maximum transient thermal impedance for MOSFET.

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